

## Low On Resistance/ Low Voltage 1 Ch 500 mA/ 1.0 A Alternative LDO

No. EA-274-211102

### OUTLINE

The RP115x is a CMOS-based positive voltage regulator featuring 500 mA/ 1.0 A that provides high ripple rejection, low dropout voltage, high output voltage accuracy, and low supply current. Internally, it consists of a voltage reference unit, an error amplifier, a resistor-net for output voltage setting, a current limit circuit, a thermal shutdown circuit, and a reverse current protection circuit. The RP115x uses a CMOS process for achieving low supply current, low On Resistance for low dropout voltage (Typ. 0.195 V (DFN1216-8,  $I_{OUT} = 1.0$  A,  $V_{SET} = 1.2$  V)) and CE function for long battery life. Excellent ripple rejection, input transient response, and load transient response make the RP115x ideal for the power sources of mobile communication equipment.

The RP115x is available in the DFN1216-8 package for space saving and the SOT-89-5 (Output Current: 1.0 A fixed) package for higher power applications. The RP115L (DFN1216-8) can choose the output current limit between 1.0 A or 500 mA by alternating the LCON pin between "H" or "L". The RP115H (SOT-89-5) can output only 1.0 A since it does not include the LCON pin.

### FEATURES

- Supply Current ..... Typ. 110  $\mu$ A
- Supply Current (Standby Mode) ..... Typ. 0.5  $\mu$ A
- Dropout Voltage ..... Typ. 0.195 V (DFN1216-8:  $I_{OUT} = 1.0$  A,  $V_{SET} = 1.2$  V)  
Typ. 0.235 V (SOT-89-5:  $I_{OUT} = 1.0$  A,  $V_{SET} = 1.2$  V)
- Ripple Rejection ..... Typ. 80 dB ( $f = 1$  kHz,  $V_{SET} \leq 1.8$  V)  
Typ. 75 dB ( $f = 1$  kHz,  $V_{SET} > 1.8$  V)
- Output Voltage Accuracy .....  $\pm 1.0\%$  ( $V_{SET} \geq 1.75$  V)
- Output Voltage Temperature Coefficient ..... Typ.  $\pm 30$  ppm/  $^{\circ}$ C ( $V_{SET} \geq 1.75$  V)
- Line Regulation ..... Typ. 0.02%/V
- Package ..... DFN1216-8, SOT-89-5
- Output Voltage Range ..... 0.7 V to 4.3 V with a 0.1-V step<sup>(1)</sup>
- Built-in Short Current Limit Circuit ..... Typ. 60 mA (DFN1216-8: LCON = "L")
- Built-in Peak Current Limit
- Built-in Thermal Shutdown Circuit ..... Thermal Shutdown Temperature: 165 $^{\circ}$ C
- Built-in Constant Slope Circuit for Start-up
- Built-in Inrush Current Suppression Circuit ..... Typ. 300 mA (DFN1216-8: LCON = "L")
- Reverse Current Protection
- Recommended Ceramic Capacitors ..... 1.0  $\mu$ F or more

<sup>(1)</sup> For the output voltage with a 0.05-V step, refer to *SELECTION GUIDE*.

## APPLICATIONS

- Portable Communication Equipment
- Electrical Appliances such as Cameras, VCRs and Camcorders
- Battery-powered Equipment
- Home Appliances, Printers, Scanners, Office Equipment Machines

## SELECTION GUIDE

The package type, the set output voltage and the auto-discharge<sup>(1)</sup> are user-selectable options.

### Selection Guide

Product Name	Package	Quantity per Reel	Pb Free	Halogen Free
RP115Lxx1*-E2	DFN1216-8	5,000 pcs	Yes	Yes
RP115Hxx1*-T1-FE	SOT-89-5	1,000 pcs	Yes	Yes

xx: Specify the set output voltage ( $V_{SET}$ ) within the range of 0.7 V to 4.3 V with a 0.1 V step.

Specify  $V_{SET}$  with a 0.05 V step as follows:

0.75 V: RP115x071\*5

1.15 V: RP115x111\*5

1.25 V: RP115x121\*5

1.35 V: RP115x131\*5

1.75 V: RP115x171\*5

1.85 V: RP115x181\*5

2.15 V: RP115x211\*5

2.75 V: RP115x271\*5

2.85 V: RP115x281\*5

2.95 V: RP115x291\*5

\*: Specify the CE pin polarity and the auto-discharge.

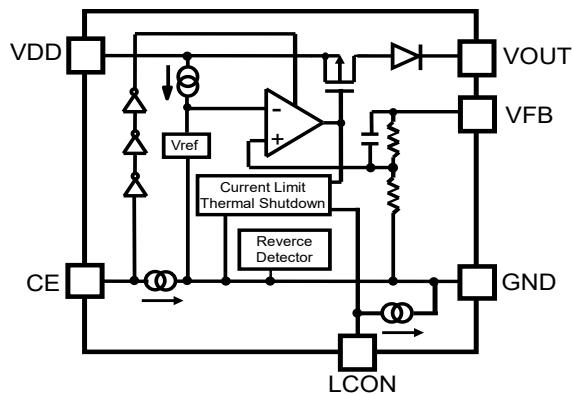
B: CE = Active-high, auto-discharge not included

D: CE = Active-high, auto-discharge included

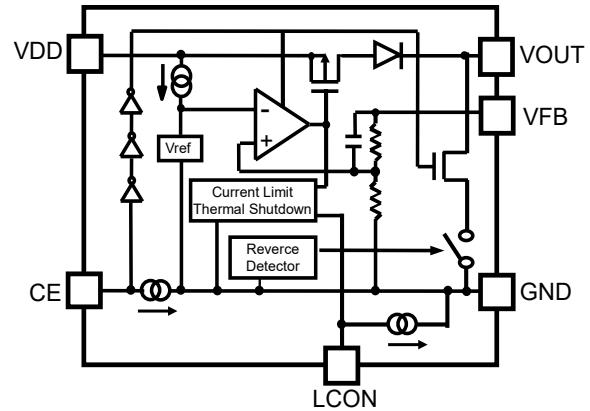
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<sup>(1)</sup> Auto-discharge function quickly lowers the output voltage to 0 V by releasing the electrical charge in the external capacitor when the chip enable signal is switched from the active mode to the standby mode.

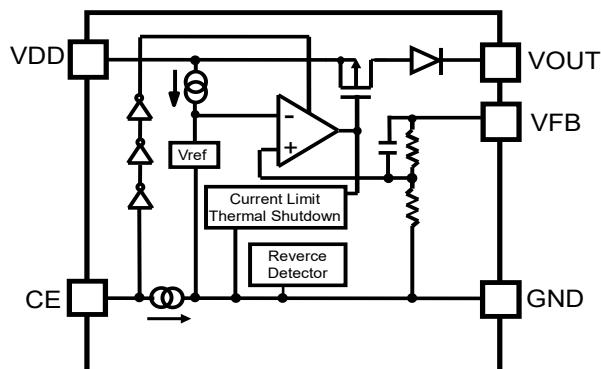
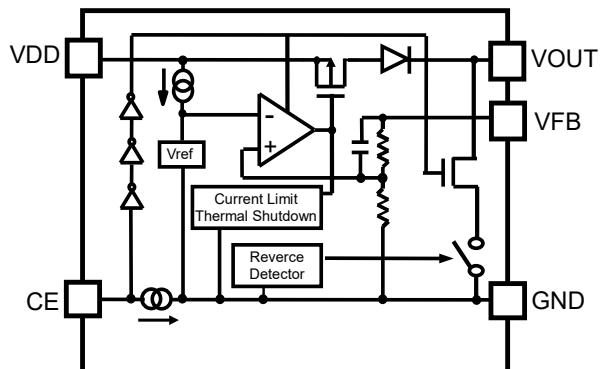
## BLOCK DIAGRAMS



RP115Lxx1B Block Diagram



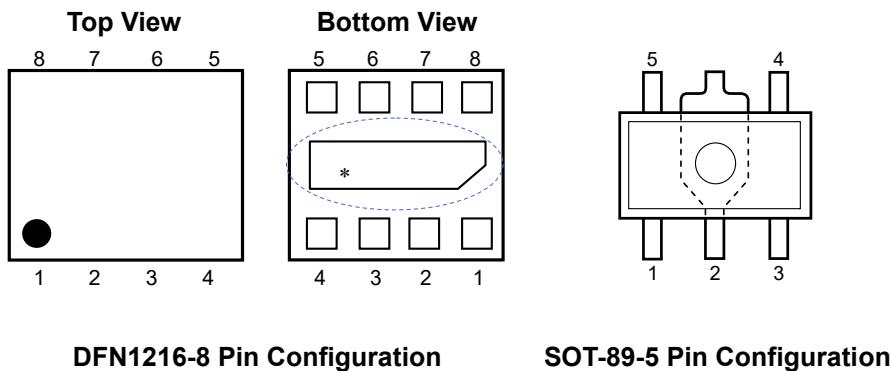
RP115Lxx1D Block Diagram

RP115Hxx1B<sup>(1)</sup> Block Diagram

RP115Hxx1D Block Diagram

<sup>(1)</sup> The RP115H does not have the LCON pin, so the output current limit is fixed at 1 A.

## PIN DESCRIPTION



### RP115L: DFN1216-8

Pin No	Symbol	Pin Description
1	VOUT <sup>(1)</sup>	Output Pin
2	VOUT <sup>(1)</sup>	Output Pin
3	LCON	Output Current Limit Alternate Pin
4	VFB <sup>(1)</sup>	Feedback Pin
5	GND	Ground Pin
6	CE	Chip Enable Pin
7	VDD <sup>(2)</sup>	Input Pin
8	VDD <sup>(2)</sup>	Input Pin

\* The tab on the bottom of the package shown by blue circle is a substrate potential (GND). It is recommended that this tab be connected to the ground plane on the board but it is possible to leave the tab floating.

### RP115H<sup>(3)</sup>: SOT-89-5

Pin No	Symbol	Pin Description
1	VFB <sup>(1)</sup>	Feedback Pin
2	GND	Ground Pin
3	CE	Chip Enable Pin
4	VDD	Input Pin
5	VOUT <sup>(1)</sup>	Output Pin

<sup>(1)</sup> The VOUT pin and the VFB pin must be wired together when mounting on the board.

<sup>(2)</sup> The VDD pins must be wired together when mounting on the board.

<sup>(3)</sup> The output current limit is fixed at 1 A.

## ABSOLUTE MAXIMUM RATINGS

### Absolute Maximum Ratings

Symbol	Parameter			Rating	Unit
$V_{IN}$	Input Voltage			6.0	V
$V_{CE}$	Input Voltage (CE Pin)			-0.3 to 6.0	V
$V_{LCON}$	Input Voltage (LCON Pin)			-0.3 to 6.0	V
$V_{out}$	Output Voltage			-0.3 to 6.0	V
$P_D$	Power Dissipation <sup>(1)</sup>	DFN1216-8	JEDEC STD. 51	1700	mW
		SOT-89-5	JEDEC STD. 51	2600	
$T_j$	Junction Temperature Range			-40 to 125	°C
$T_{stg}$	Storage Temperature Range			-55 to 125	°C

### ABSOLUTE MAXIMUM RATINGS

Electronic and mechanical stress momentarily exceeded absolute maximum ratings may cause permanent damage and may degrade the life time and safety for both device and system using the device in the field. The functional operation at or over these absolute maximum ratings is not assured.

## RECOMMENDED OPERATING CONDITIONS

### Recommended Operating Conditions

Symbol	Item	Rating	Unit
$V_{IN}$	Input Voltage <sup>(2)</sup>	1.4 to 5.25	V
$T_a$	Operating Temperature Range	-40 to 85	°C

### RECOMMENDED OPERATING CONDITIONS

All of electronic equipment should be designed that the mounted semiconductor devices operate within the recommended operating conditions. The semiconductor devices cannot operate normally over the recommended operating conditions, even if they are used over such conditions by momentary electronic noise or surge. And the semiconductor devices may receive serious damage when they continue to operate over the recommended operating conditions.

<sup>(1)</sup> Refer to *POWER DISSIPATION* for detailed information.

<sup>(2)</sup> In case of operating the device beyond 5.25 V, do not exceed 5.5 V with 500 total operating hours.

## ELECTRICAL CHARACTERISTICS

$V_{IN} = V_{SET}^{(1)} + 1.0 \text{ V}$ ,  $I_{OUT} = 1 \text{ mA}$ ,  $C_{IN} = C_{OUT} = 1.0 \mu\text{F}$ , unless otherwise noted.

The specifications surrounded by   are guaranteed by design engineering at  $-40^\circ\text{C} \leq Ta \leq 85^\circ\text{C}$ .

RP115x Electrical Characteristics <span style="float: right;"><math>(Ta = 25^\circ\text{C})</math></span>							
Symbol	Item	Conditions		Min.	Typ.	Max.	Unit
$V_{OUT}$	Output Voltage	$Ta = 25^\circ\text{C}$	$V_{SET} \geq 1.75 \text{ V}$	x0.99		x1.01	V
			$V_{SET} < 1.75 \text{ V}$	-18		+18	mV
		$-40^\circ\text{C} \leq Ta \leq 85^\circ\text{C}$	$V_{SET} \geq 1.75 \text{ V}$	x0.985		x1.015	V
			$V_{SET} < 1.75 \text{ V}$	Refer to Set Output Voltage-specific Output Voltage Characteristics.			
$I_{LIM}$	Output Current Limit	$V_{IN} = V_{SET} + 0.5 \text{ V}$	$LCON = "L"$	500			mA
			$LCON = "H"^{(2)}$	1.0			A
$\Delta V_{OUT} / \Delta I_{OUT}$	Load Regulation	$V_{IN} = V_{SET} + 0.5 \text{ V}$ $1 \text{ mA} \leq I_{OUT} \leq 500 \text{ mA}$	$LCON = "L"$	1		20	mV
		$V_{IN} = V_{SET} + 0.5 \text{ V}$ $1 \text{ mA} \leq I_{OUT} \leq 1.0 \text{ A}$	$LCON = "H"^{(2)}$			40	
$V_{DIF}$	Dropout Voltage	Refer to Set Output Voltage-specific Dropout Voltage Characteristics.					
$I_{SS}$	Supply Current	$I_{OUT} = 0 \text{ mA}$		110	160		$\mu\text{A}$
Istandby	Standby Current	$V_{CE} = 0 \text{ V}$		0.5	3.0		$\mu\text{A}$
$\Delta V_{OUT} / \Delta V_{IN}$	Line Regulation	$V_{SET} + 0.5 \text{ V} \leq V_{IN} \leq 5.25 \text{ V}$ ( $V_{IN} \geq 1.4 \text{ V}$ )		0.02	0.10		%/V
RR	Ripple Rejection	$f = 1 \text{ kHz}$ , Ripple 0.2 Vp-p, $V_{IN} = V_{SET} + 1.0 \text{ V}$ , $I_{OUT} = 30 \text{ mA}$	$V_{SET} > 1.8 \text{ V}$		75		dB
			$V_{SET} \leq 1.8 \text{ V}$		80		dB
$\Delta V_{OUT} / \Delta T_a$	Output Voltage Temperature Coefficient	$-40^\circ\text{C} \leq Ta \leq 85^\circ\text{C}$	$V_{SET} \geq 1.75 \text{ V}$		±30		ppm/ $^\circ\text{C}$
			$V_{SET} < 1.75 \text{ V}$		±100		
$I_{SC}$	Short Current Limit	$V_{OUT} = 0 \text{ V}^{(3)}$	$LCON = "L"$		60		mA
			$LCON = "H"^{(2)}$		110		
$I_{CE}$	CE Pull-down Current			0.05	0.3	0.6	$\mu\text{A}$
$V_{CEH}$	CE Input Voltage "H"			1.0			V
$V_{CEL}$	CE Input Voltage "L"					0.4	V
$I_{LCON}$	LCON Pull-down Current (RP115L only)			0.05	0.3	0.6	$\mu\text{A}$

(1)  $V_{SET}$ : Set Output Voltage

(2) The electrical characteristics of the RP115H is as same as when  $LCON = "H"$ .

(3) The short current limit is the value when the VOUT pin is short circuited to GND after the device is completely started up. The inrush current flows when the VOUT pin is short circuited to GND while the VOUT pin is short-circuited to GND before the device is completely started up.

## ELECTRICAL CHARACTERISTICS (continued)

$V_{IN} = V_{SET}^{(1)} + 1.0 \text{ V}$ ,  $I_{OUT} = 1 \text{ mA}$ ,  $C_{IN} = C_{OUT} = 1.0 \mu\text{F}$ , unless otherwise noted.

The specifications surrounded by   are guaranteed by design engineering at  $-40^\circ\text{C} \leq Ta \leq 85^\circ\text{C}$ .

RP115x Electrical Characteristics						$(Ta = 25^\circ\text{C})$	
Symbol	Item	Conditions		Min.	Typ.	Max.	Unit
$V_{LCONH}$	LCON Input Voltage "H" (RP115L only)			1.0			V
$V_{LCONL}$	LCON Input Voltage "L" (RP115L only)				0.4		V
$T_{TSD}$	Thermal Shutdown Temperature Threshold	$T_j$ , Rising			165		$^\circ\text{C}$
$T_{TSR}$		$T_j$ , Falling			110		$^\circ\text{C}$
$I_{REV}$	Reverse Current	$V_{OUT} = V_{SET} + 1.0 \text{ V}$ $0 \leq V_{IN} \leq V_{OUT}$	$V_{SET} \geq 1.75 \text{ V}$		7.5		$\mu\text{A}$
			$V_{SET} < 1.75 \text{ V}$		10		
$V_{REV\_DET}^{(2)}$	Detection Offset Voltage in Reverse Current Protection Mode <sup>(3)</sup>	$V_{OUT} \geq 0.7 \text{ V}$ , $0 \leq V_{IN} \leq 5.25 \text{ V}$			20		mV
$V_{REV\_REL}^{(4)}$	Release Offset Voltage in Reverse Current Protection Mode <sup>(3)</sup>	$V_{OUT} \geq 0.7 \text{ V}$ , $0 \leq V_{IN} \leq 5.25 \text{ V}$			30	50	mV
en	Output Noise	BW = 10 Hz to 100 kHz	$V_{SET} \geq 1.75 \text{ V}$		17 $\times V_{SET}$		$\mu\text{VRms}$
			$V_{SET} < 1.75 \text{ V}$		35 $\times V_{SET}$		
$R_{LOW}$	Auto-discharge Nch Tr. On-resistance (RP115xx1D only)	$V_{IN} = 4.0 \text{ V}$ , $V_{CE} = 0 \text{ V}$			60		$\Omega$
$I_{RUSH}$	Inrush Current Limit	CC mode <sup>(5)</sup>	$LCON = "L"$		300		$\text{mA}$
			$LCON = "H"^{(6)}$		500		

All test items listed under *Electrical Characteristics* are done under the pulse load condition ( $T_j \approx Ta = 25^\circ\text{C}$ ) except Output Noise, Ripple Rejection, and Output Voltage Temperature Coefficient.

<sup>(1)</sup>  $V_{SET}$ : Set Output Voltage

<sup>(2)</sup>  $V_{REV\_DET} = V_{IN} - V_{OUT}$

<sup>(3)</sup> The guaranteed operating voltage range of the reverse current protection circuit is  $V_{OUT} \geq 0.7 \text{ V}$ . When  $V_{IN} = V_{OUT} = 0 \text{ V}$ , the reverse current protection mode is constantly active.

<sup>(4)</sup>  $V_{REV\_REL} = V_{IN} - V_{OUT}$

<sup>(5)</sup> For detailed information, refer to *Start-up Characteristics Using Constant Slope Circuit*.

<sup>(6)</sup> The electrical characteristics of the RP115H is as same as when  $LCON = "H"$ .

### Set Output Voltage-specific Output Voltage Characteristics

Set Output Voltage $V_{SET}$ (V)	Output Voltage $V_{OUT}$ (mV)	
	Min.	Max.
0.7	-33	+28
0.8	-35	+29
0.9	-37	+30
1.0	-39	+31
1.1	-41	+33
1.2	-43	+34
1.3	-45	+35
1.4	-47	+36
1.5	-49	+38
1.6	-51	+39
1.7	-53	+40

### Set Output Voltage-specific Dropout Voltage Characteristics

RP115L: DFN1216-8

(Ta = 25°C)

Set Output Voltage $V_{SET}$ (V)	Dropout Voltage $V_{DIF}$ (V)					
	RP115L				RP115H	
	$I_{OUT} = 500$ mA		$I_{OUT} = 1000$ mA		$I_{OUT} = 1000$ mA	
	Typ.	Max.	Typ.	Max.	Typ.	Max.
0.7 ≤ $V_{SET}$ < 1.1	*	*	*	*	*	*
1.1 ≤ $V_{SET}$ < 1.2	*	*	*	0.300	*	0.350
1.2 ≤ $V_{SET}$ < 1.3	*	*	0.195	0.275	0.235	0.330
1.3 ≤ $V_{SET}$ < 1.5	0.095	0.135	0.185	0.260	0.225	0.320
1.5 ≤ $V_{SET}$ < 1.75	0.085	0.120	0.165	0.235	0.205	0.295
1.75 ≤ $V_{SET}$ < 2.6	0.075	0.110	0.150	0.215	0.190	0.270
2.6 ≤ $V_{SET}$ < 3.3	0.065	0.090	0.130	0.180	0.170	0.240
3.3 ≤ $V_{SET}$ ≤ 4.3	0.060	0.085	0.125	0.170	0.165	0.225

If the dropout voltage falls below the release offset value of reverse current protection mode ( $V_{REV\_REL}$ ), the reverse current protection circuit may repeat the detection and release operations. Please refer to Reverse Current Protection Circuit.

\* Input voltage should be equal or more than the minimum operating voltage (1.4 V).

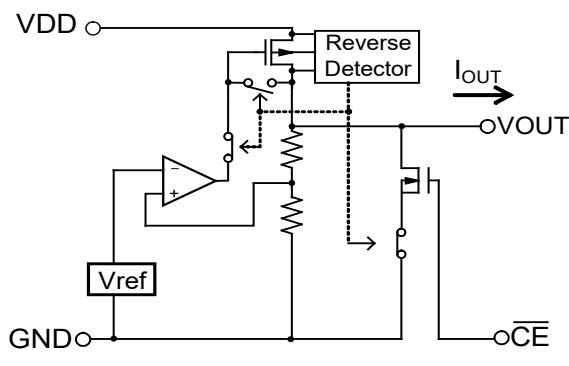
## THEORY OF OPERATION

### Reverse Current Protection Circuit

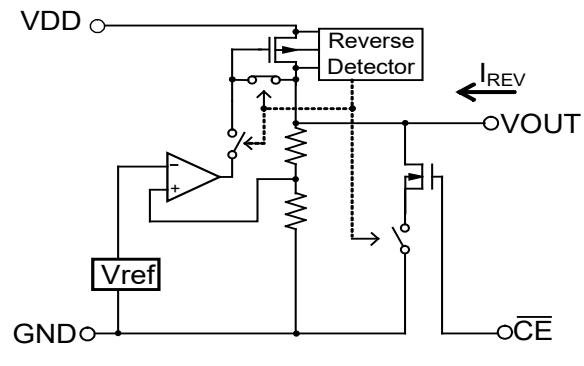
The RP115x includes a Reverse Current Protection Circuit, which stops the reverse current from  $V_{OUT}$  pin to  $V_{DD}$  pin or to GND pin when  $V_{OUT}$  becomes higher than  $V_{IN}$ . Usually, the LDO using Pch output transistor contains a parasitic diode between  $V_{DD}$  pin and  $V_{OUT}$  pin. Therefore, if  $V_{OUT}$  is higher than  $V_{IN}$ , the parasitic diode becomes forward direction. As a result, the current flows from  $V_{OUT}$  pin to  $V_{DD}$  pin. The RP115x switches the mode to the reverse current protection mode before  $V_{IN}$  becomes lower than  $V_{OUT}$  by connecting the parasitic diode of Pch output transistor to the backward direction, and connecting the gate to  $V_{OUT}$  pin. As a result, the Pch output transistor is turned off. However, from  $V_{OUT}$  pin to GND pin, via the internal divider resistors, very small current  $I_{REV}$  flows.

Switching to either the normal mode or to the reverse current protection mode is determined by the magnitude of  $V_{IN}$  voltage and  $V_{OUT}$  voltage. For the stable operation, offset and hysteresis are set as the threshold. The detector threshold is set to  $V_{REV\_DET}$  and the released voltage is set to  $V_{REV\_REL}$ . Therefore, the minimum dropout voltage under the small load current condition is restricted by the value of  $V_{REV\_REL}$ .

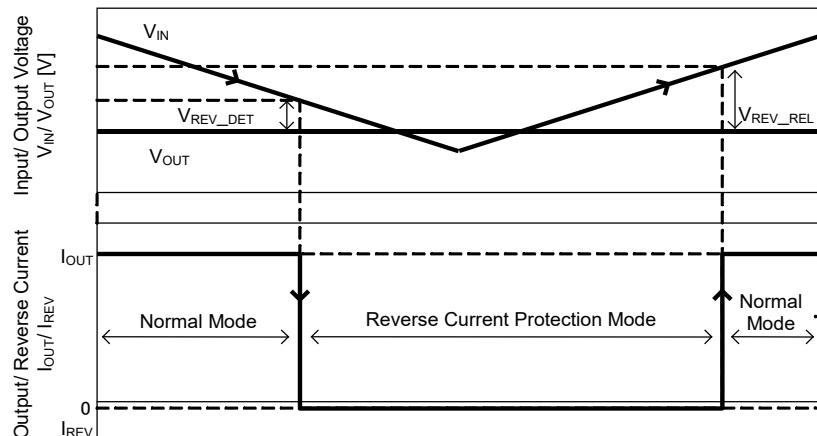
Following figures show the diagrams of each mode, and the load characteristics of each mode. When giving the  $V_{OUT}$  pin a constant-voltage and decreasing the  $V_{IN}$  voltage, the dropout voltage will become lower than  $V_{REV\_DET}$ . As a result, the reverse current protection starts to function to stop the load current. By increasing the dropout voltage higher than  $V_{REV\_REL}$ , the protection mode will be released to let the load current to flow. If the dropout voltage to be used is lower than  $V_{REV\_REL}$ , the detection and the release may be repeated. The operating voltage guaranteed level of the reverse current protection circuit is for  $V_{OUT} \geq 0.7V$ . If  $V_{IN}=0V$ , the reverse current protection mode becomes always active.



Normal Operation Mode



Reverse Current Protection Mode

**Detection/ Release Timing of Reverse Current Protection**

### Start-up Characteristics Using Constant Slope Circuit

Constant slope circuit is included in the RP115x to prevent the overshoot of the output voltage. The start-up time ( $t_{ON}$ ) is 100  $\mu$ s (Typ.). If inrush current increases due to the large capacitance of  $C_{OUT}$ , the operation mode will be shifted from Constant Slope (CS) mode to Constant Current (CC) mode. The CC mode maintains a constant level of inrush current. In the CC mode,  $t_{ON}$  varies according to the size of  $C_{OUT}$  and the amount of load current.

### Start-up Time and Inrush Current Estimations

Start-up time and inrush current in the CS mode and the CC mode can be estimated as follows.

#### [CS Mode]

Start-up Time:  $t_{ON} = 100 \mu\text{s}$  (Typ.)

Inrush Current:  $I_{RUSH} = C_{OUT} \cdot V_{SET} / t_{ON} + I_{OUT}^{(1)}$

Note: If the result of the above calculation is more than the following values, the operation mode will be shifted from the CS mode to the CC mode.

LCON = "L" ..... 300 mA (Typ.)

LCON = "H" ..... 500 mA (Typ.)

#### [CC Mode]

Start-up Time:  $t_{ON} = C_{OUT} \cdot V_{SET} / I_{CO}^{(2)}$

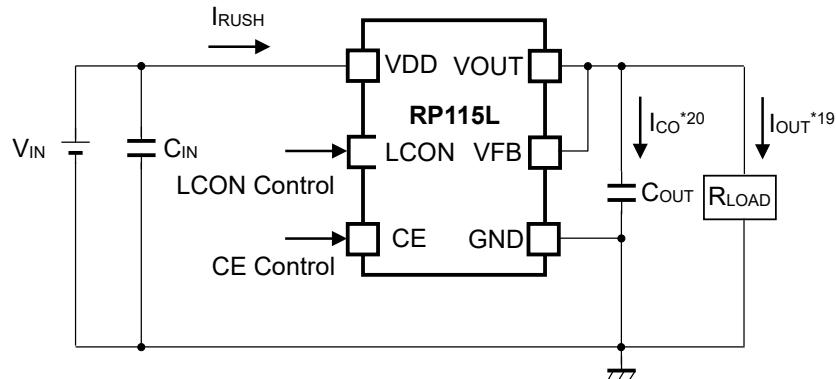
Inrush Current:  $I_{RUSH}$

LCON = "L" ..... 300 mA (Typ.)

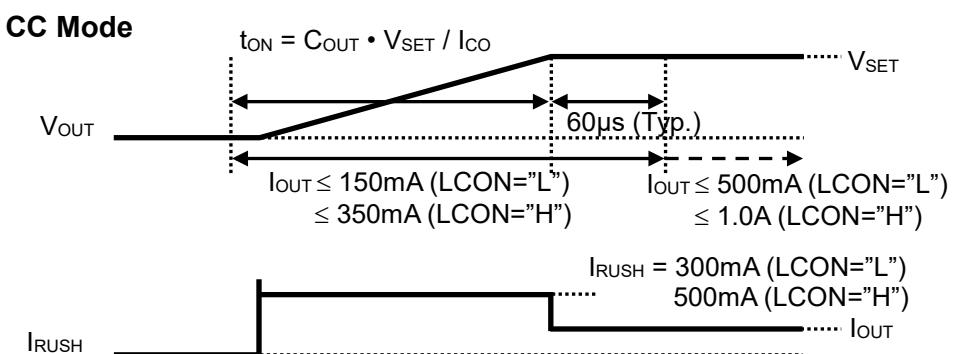
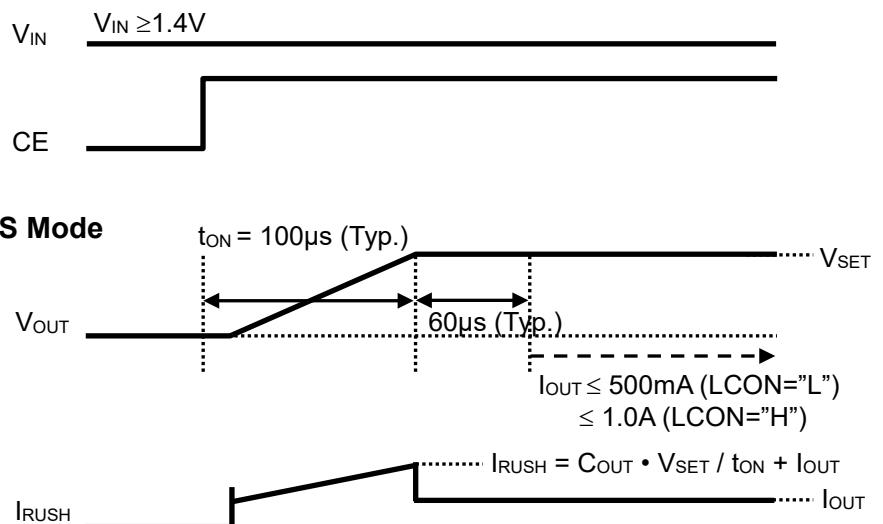
LCON = "H" ..... 500 mA (Typ.)

<sup>(1)</sup>  $I_{OUT}$ : When  $R_{LOAD}$  is connected to load,  $I_{OUT}$  can be calculated by  $R_{LOAD} = V_{SET} / I_{OUT}$ .

<sup>(2)</sup>  $I_{CO}$ :  $I_{CO}$  is a charge current of  $C_{OUT}$  and can be calculated roughly by  $I_{RUSH} \approx I_{CO} + I_{OUT}$ .



Circuit Example



Timing Chart of Start-up Operation

## Precautions before Use

During the start-up, the inrush current limit circuit is in operation; therefore, the load current ( $I_{OUT}$ ) should be drawn after the output voltage ( $V_{OUT}$ ) reached the preset value (Best timing:  $t_{ON} + 60 \mu s$  or more). If the load current is drawn during the start-up, it should be within the following values.

LCON = "L" .....  $I_{OUT} \leq 150 \text{ mA}$

LCON = "H" .....  $I_{OUT} \leq 350 \text{ mA}$

In the CC mode,  $I_{RUSH}$  is limited until  $V_{OUT}$  reaches the preset value.  $I_{RUSH} \approx I_{CO} + I_{OUT}$  is true; therefore, if large  $I_{OUT}$  is drawn during the start-up, the charge current ( $I_{CO}$ ) of  $C_{OUT}$  decreases and  $t_{ON}$  becomes longer. Please refer to *Start-up Time and Inrush Current Estimations*.

In order to control the start-up operation by using the CS mode or CC mode, input "H" into the CE pin while  $V_{IN} \geq 1.4 \text{ V}$ . If "H" is input into the CE pin while  $V_{IN}$  is less than the minimum operating voltage, the operation may not be controlled by the CS mode or CC mode.

When starting up the device while the short circuit is occurring between the  $V_{OUT}$  pin and GND, the short current protection circuit does not control the current but the current limit circuit does.

When there's excessive heat generation in the device, thermal shutdown circuit shuts down the circuitry before the device overheats dangerously.

## LCON Pin (RP115L only)

By alternating the LCON pin between "H" or "L", the RP115L can choose the output current limit either 1.0 A or 500 mA. Please note that during start-up ( $t_{ON} + 60 \mu s$  (Typ.)), do not change the logic of the LCON pin.

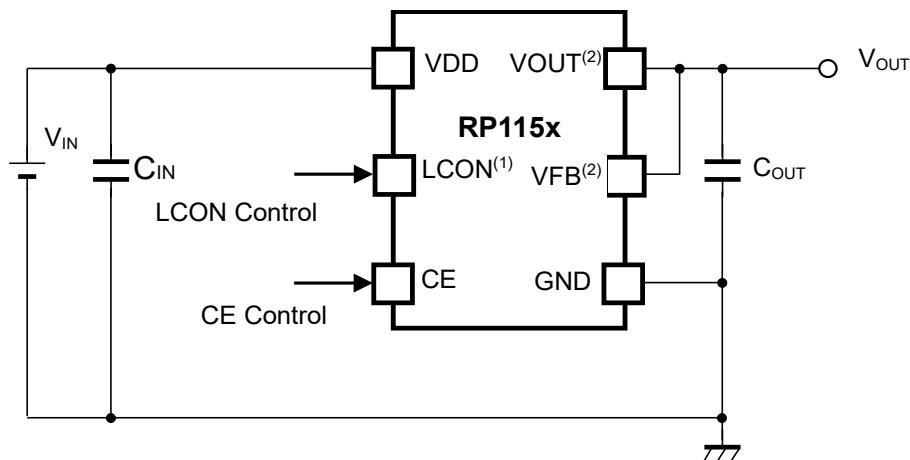
LCON = "L" ..... 500 mA

LCON = "H" ..... 1.0 A

## Application Example

Even when using the RP115L with LCON = "H",  $I_{RUSH}$  in the CC mode can be reduced from 500 mA (Typ.) to 300 mA (Typ.) by starting up the IC with LCON = "L". Please refer to *Start-up Characteristics Using Constant Slope Circuit*.

## APPLICATION INFORMATION



**RP115x Typical Application Circuit**

### External Components

Symbol	Description
C <sub>IN</sub>	1.0 $\mu$ F, Ceramic Capacitor, GRM155R61A105KE15 (MURATA)
C <sub>OUT</sub>	1.0 $\mu$ F, Ceramic Capacitor, GRM155R61A105KE15 (MURATA) 2.2 $\mu$ F, Ceramic Capacitor, GRM155R61A225KE95 (MURATA)

### Precautions When Selecting External Components

- In this device, phase compensation is provided to secure stable operation even when the load current is varied. For this purpose, use a 1.0- $\mu$ F or more output capacitor (C<sub>OUT</sub>).
- A ceramic capacitor has different temperature characteristics and bias dependencies depending on the size, manufacturer or part number of a capacitor. Careful evaluation is required. When using a 1.75-V product under the environment of -20°C or lower, choose a 2.2- $\mu$ F or more C<sub>OUT</sub>.
- In case of using a tantalum-type capacitor with a large ESR (Equivalent Series Resistance), the output might become unstable. Careful evaluation on frequency characteristics is required.

<sup>(1)</sup> The LCON pin is included in the RP115L (DFN2020-8B) only.

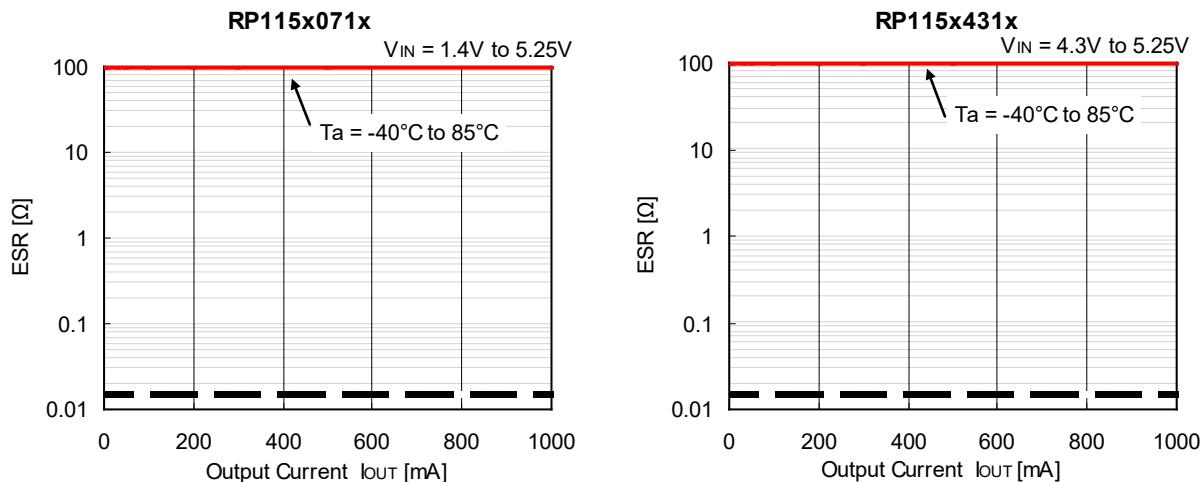
<sup>(2)</sup> Connect the VOUT and VFB pins together.

## Equivalent Series Resistance (ESR) vs. Output Current

Ceramic type output capacitor is recommended for the RP115x but any capacitor with low ESR can be used. The graphs below show the relation between  $I_{OUT}$  and ESR (noise level: average 40  $\mu$ V or less).

### Measurement Conditions

- Noise Frequency Band Width: 10 Hz to 2 MHz
- Operating Temperature Range:  $-40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$
- Hatched Area: Output noise level is average 40  $\mu$ V or less.
- $C_{IN}, C_{OUT}$ : 1.0  $\mu\text{F}$  or more



## TECHNICAL NOTES

The performance of a power source circuit using this device is highly dependent on a peripheral circuit. A peripheral component or the device mounted on PCB should not exceed its rated voltage, rated current or rated power. When designing a peripheral circuit, please be fully aware of the following points.

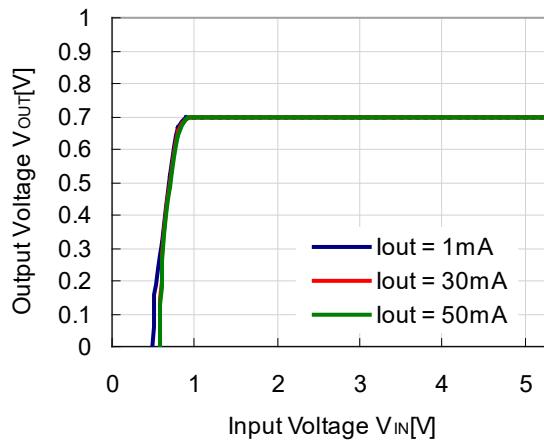
- Place the external components as close as possible to the device with shortest-distance wirings. Connect an input capacitor ( $C_{IN}$ ) between the VIN and GND pins with shortest-distance wiring.
- Ensure the VDD and GND wirings are sufficiently robust. If the impedance of wiring between the VDD and GND pins is high, it may cause noise pickup or unstable operation.
- Connect an output capacitor ( $C_{OUT}$ ) between the VOUT and GND pins with shortest-distance wiring.

## TYPICAL CHARACTERISTICS

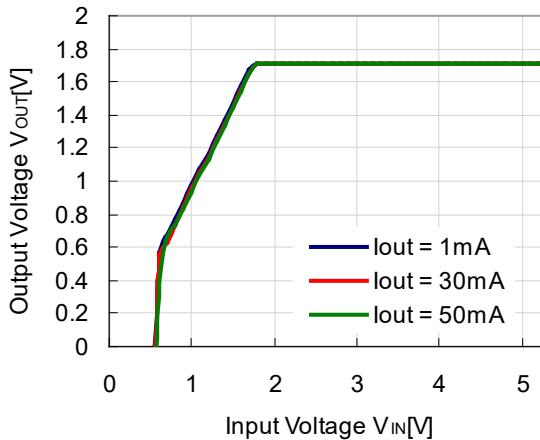
Note: Typical Characteristics are intended to be used as reference data; they are not guaranteed.

### 1) Output Voltage vs. Input Voltage ( $C_{IN} = \text{Ceramic } 1.0 \mu\text{F}$ , $C_{OUT} = \text{Ceramic } 1.0 \mu\text{F}$ , $T_a = 25^\circ\text{C}$ )

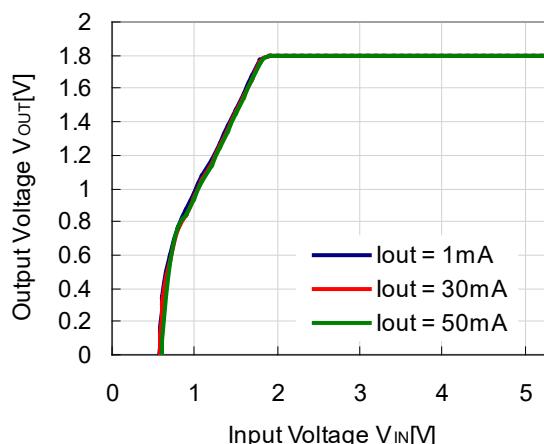
RP115x071x



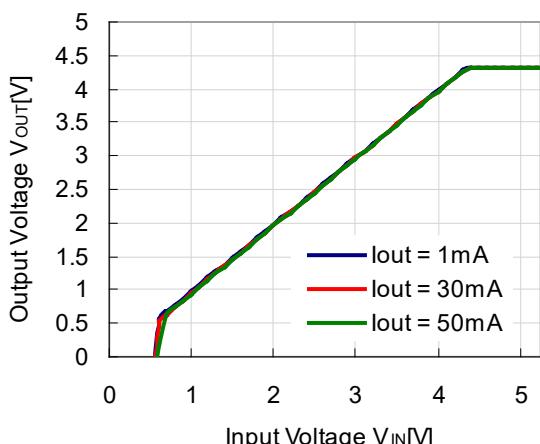
RP115x171x



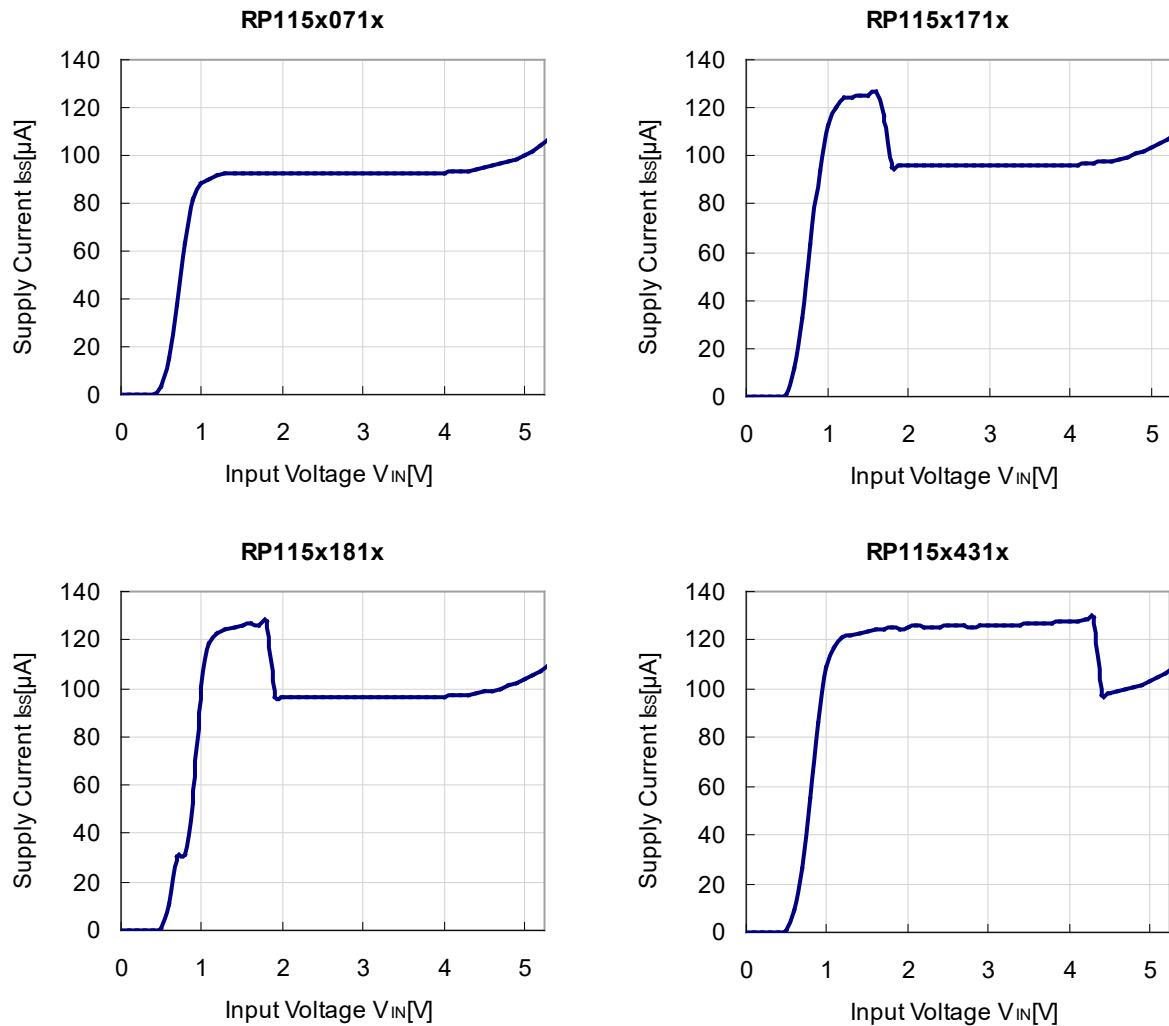
RP115x181x



RP115x431x

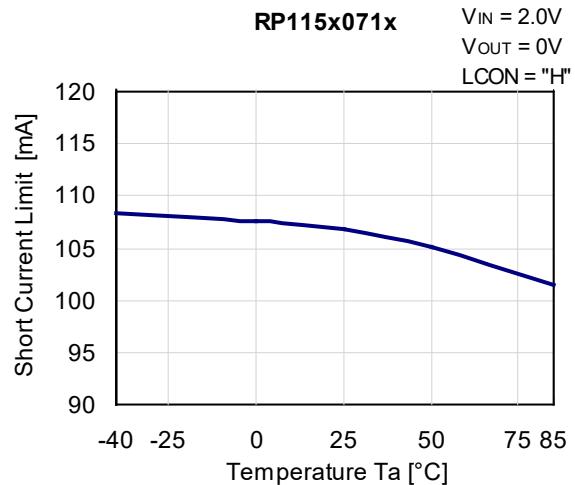
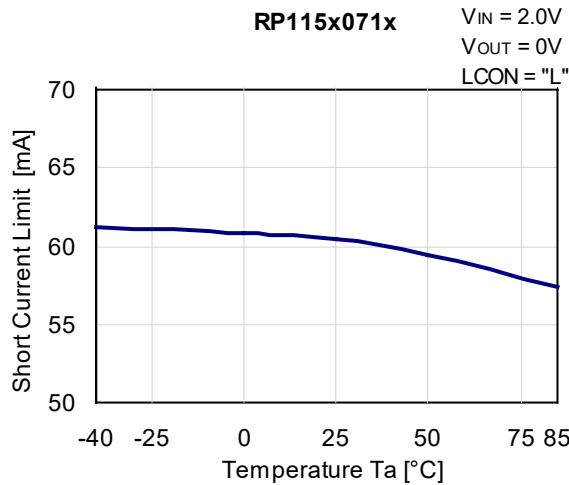
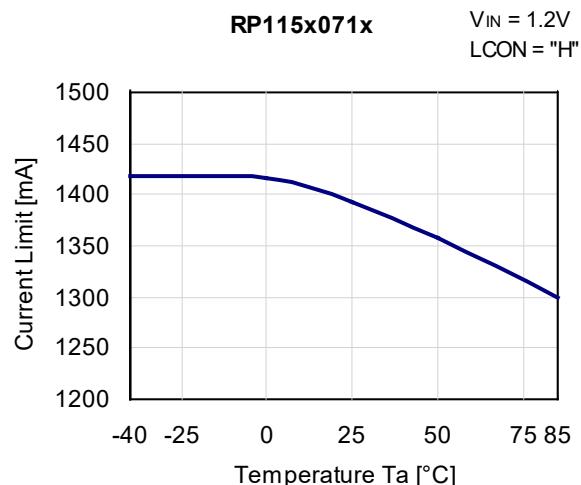
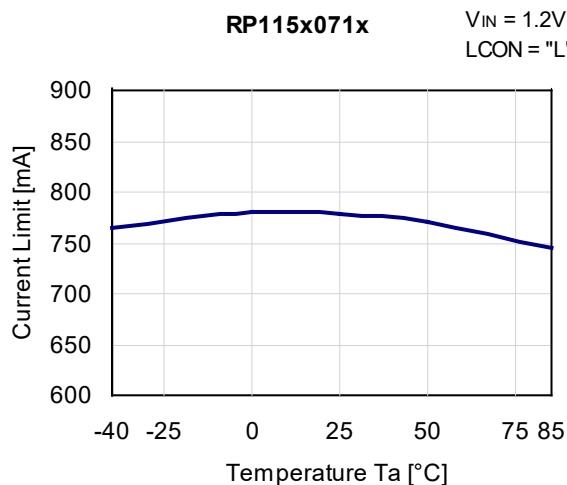


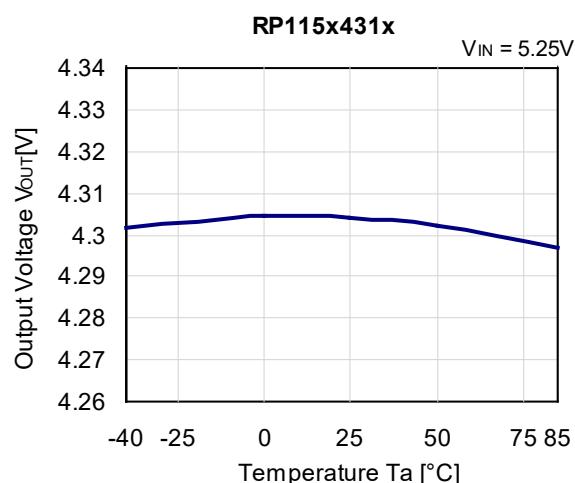
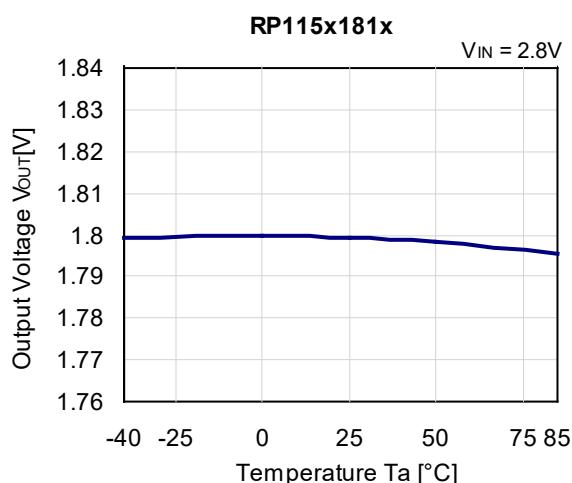
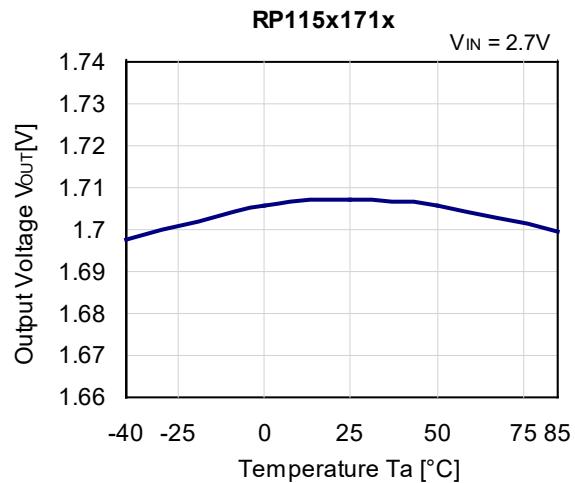
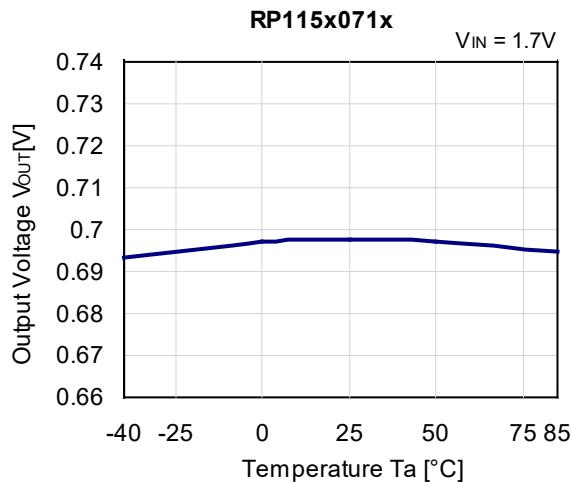
**2) Supply Current vs. Input Voltage ( $C_{IN}$  = Ceramic 1.0  $\mu$ F,  $C_{OUT}$  = Ceramic 1.0  $\mu$ F,  $T_a$  = 25°C)**

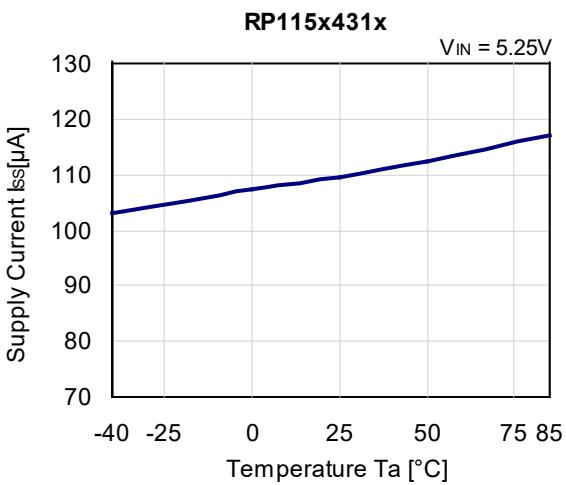
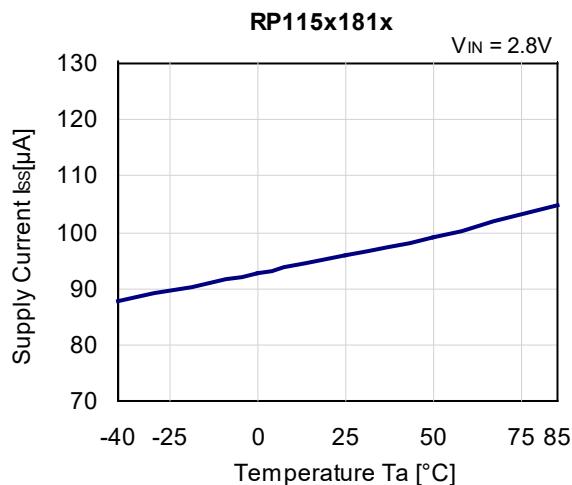
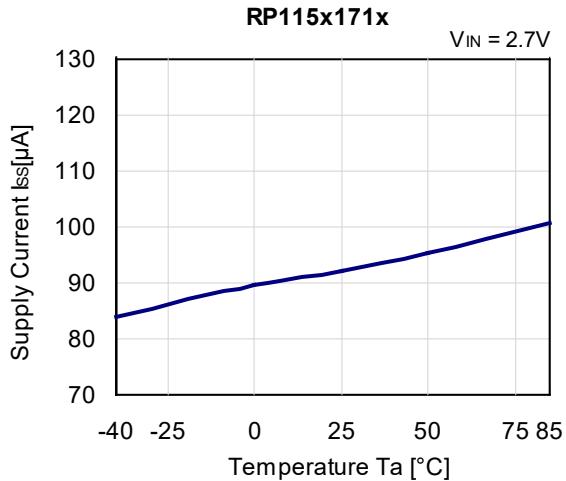
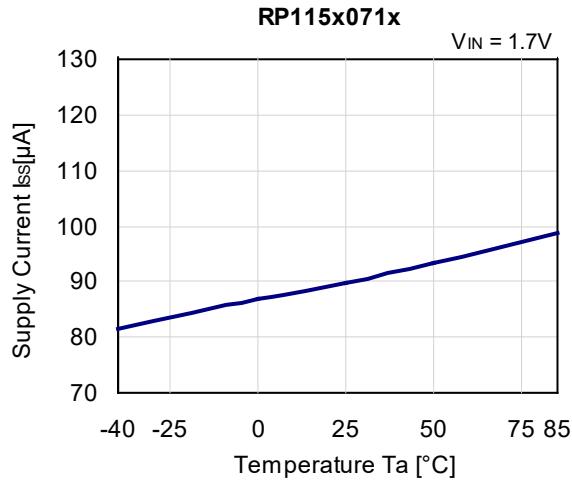


**Short Current Limit vs. Temperature and Current Limit vs. Temperature**

The RP115x contains a peak current limit circuit which protect the regulator from damage by overcurrent if the output pin (VOUT) and the ground pin (GND) are shorted. The short-circuiting causes the overheating of the device which leads a thermal shutdown circuit to operate. If the peak current limit circuit and the thermal shutdown circuit work at the same time, fold-back type dropping characteristics cannot be measured. As for the short-circuit current and the peak current limit circuit, please refer to 3) Short Current Limit vs. Temperature and 4) Current Limit vs. Temperature.

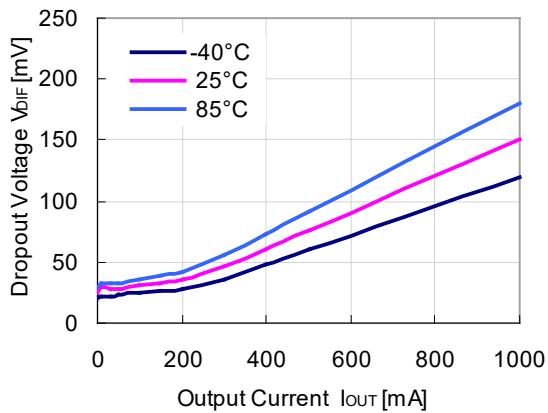
**3) Short Current Limit vs. Temperature****4) Peak Current Limit vs. Temperature**

**5) Output Voltage vs. Temperature ( $C_{IN}$  = Ceramic 1.0  $\mu$ F,  $C_{OUT}$  = Ceramic 1.0  $\mu$ F,  $I_{OUT}$  = 1 mA)**

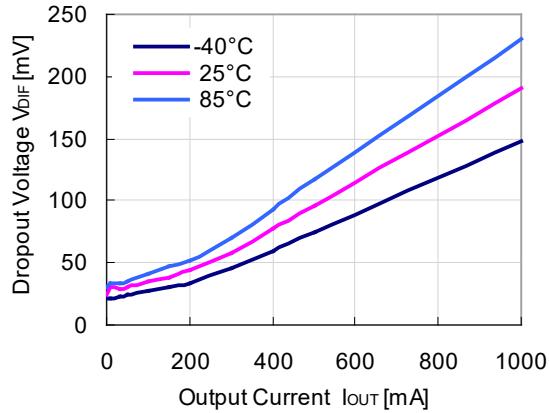
**6) Supply Current vs. Temperature ( $C_{IN}$  = Ceramic 1.0  $\mu$ F,  $C_{OUT}$  = Ceramic 1.0  $\mu$ F,  $I_{OUT}$  = 0 mA)**

7) Dropout Voltage vs. Output Current ( $C_{IN}$  = Ceramic 1.0  $\mu$ F,  $C_{OUT}$  = Ceramic 1.0  $\mu$ F)

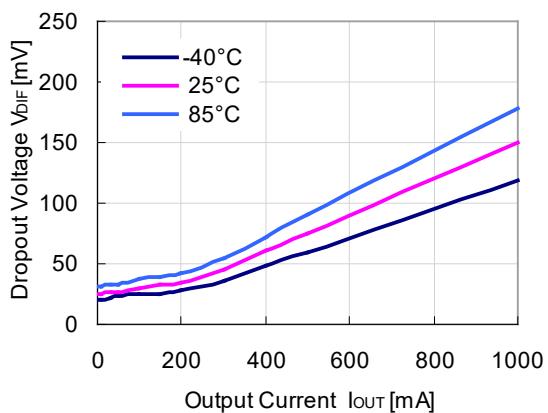
RP115L171x



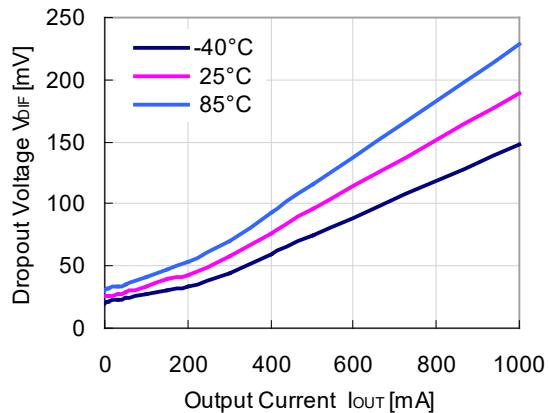
RP115H171x



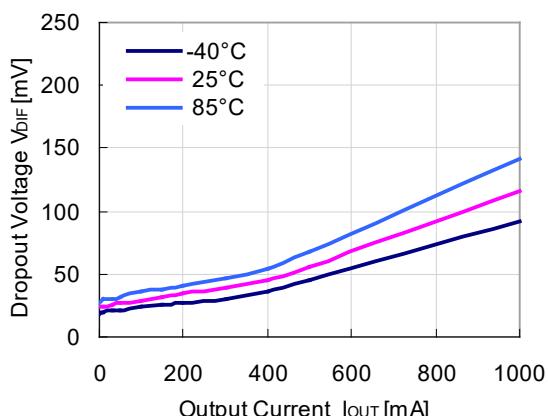
RP115L181x



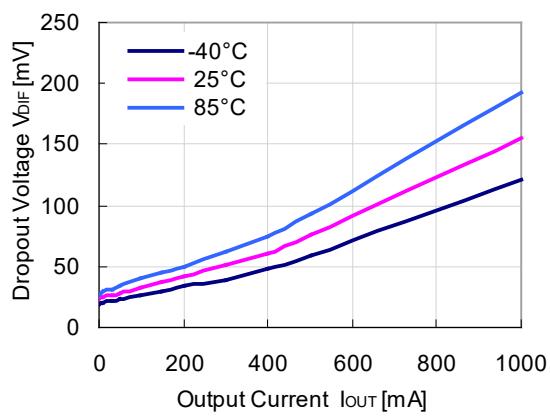
RP115H181x

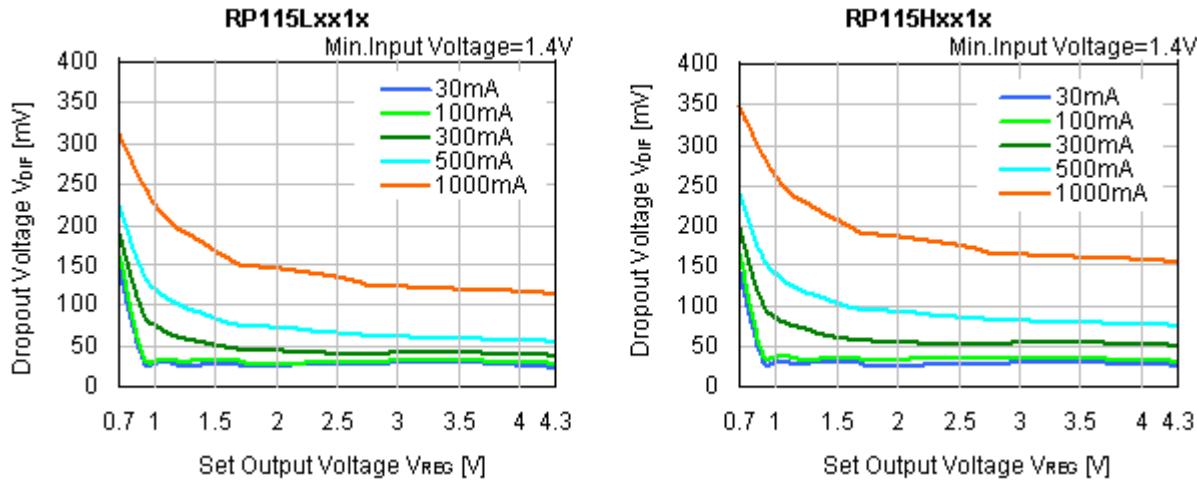
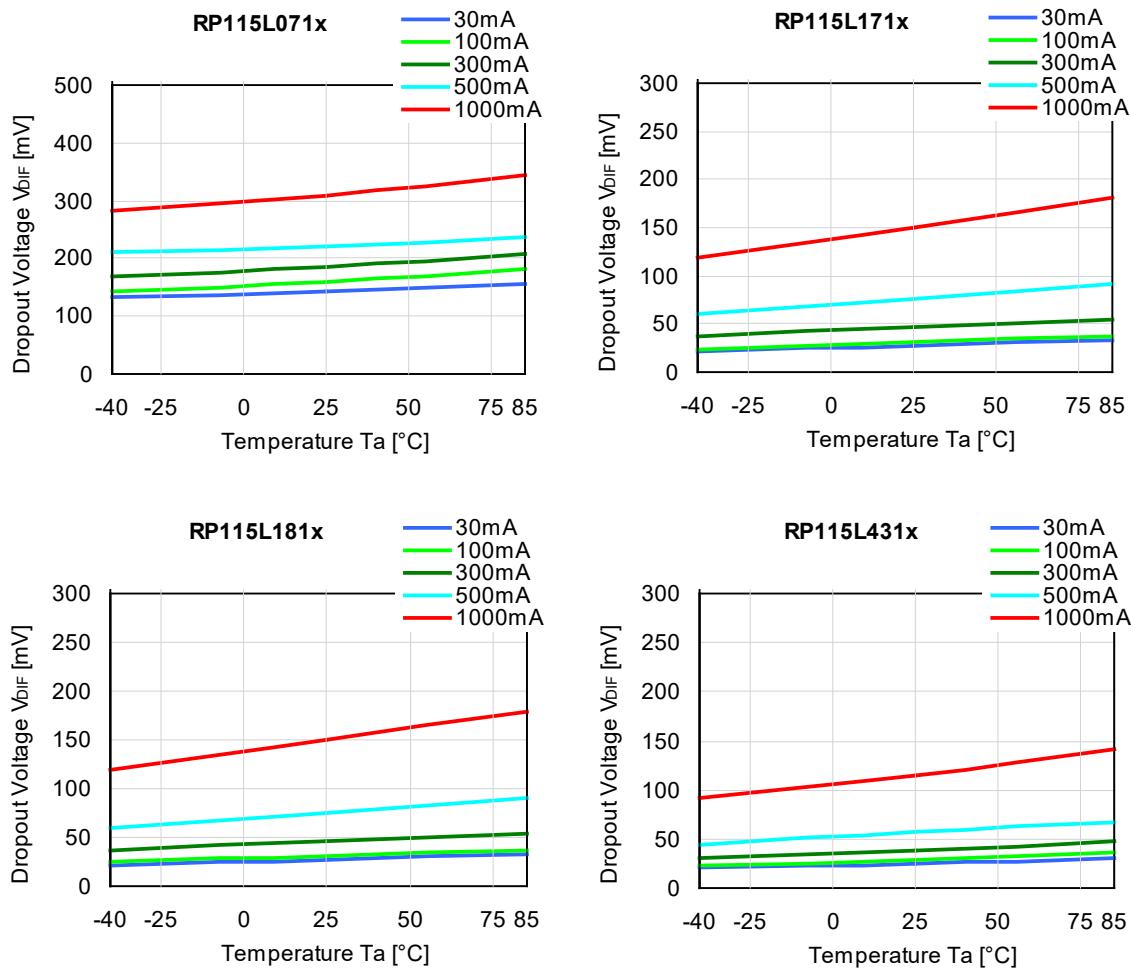


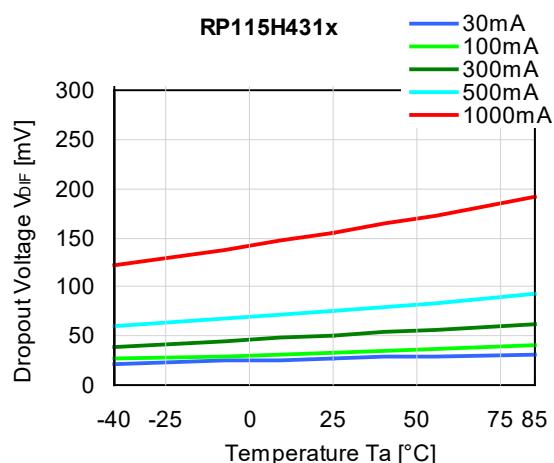
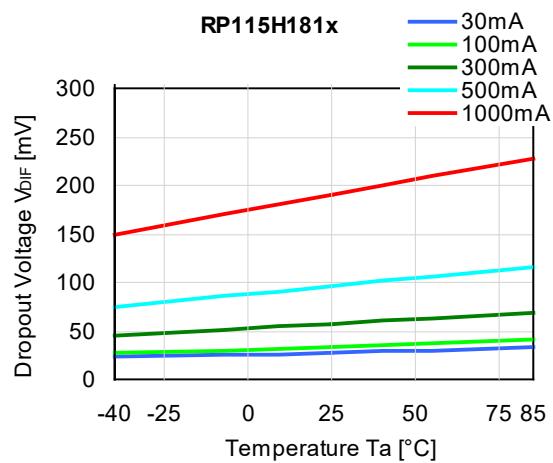
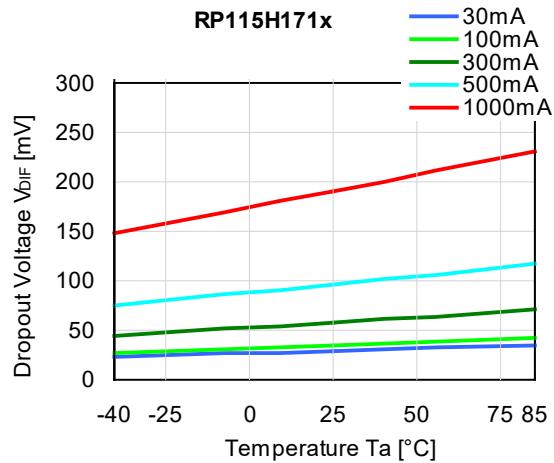
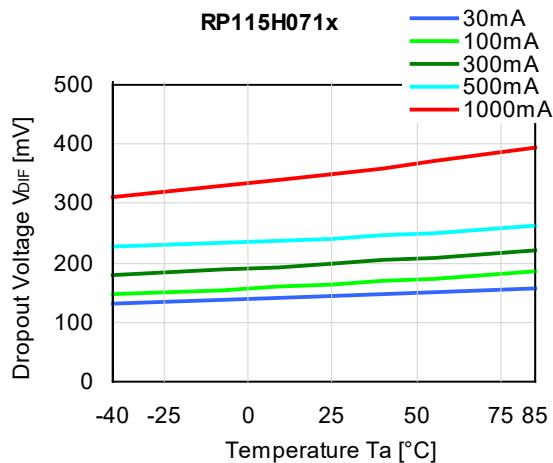
RP115L431x



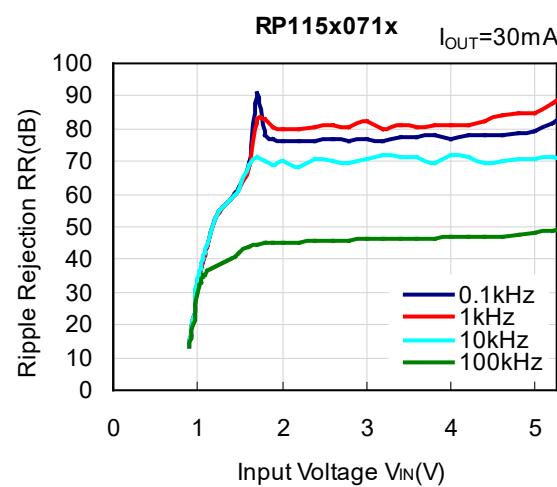
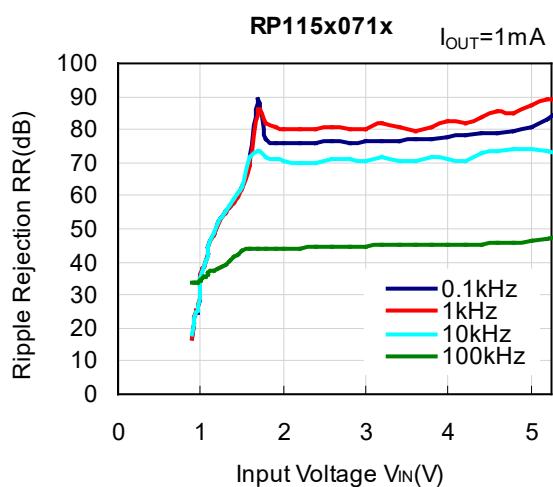
RP115H431x

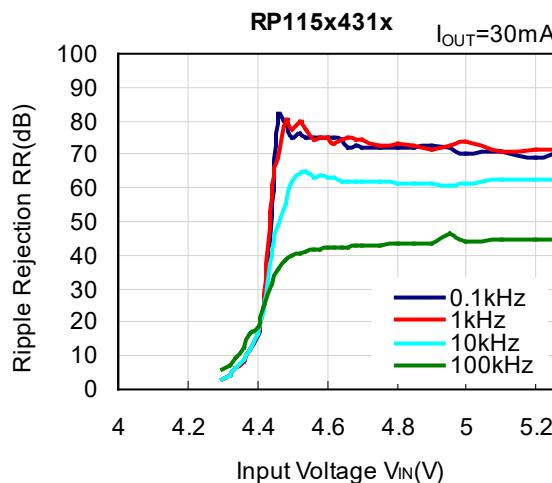
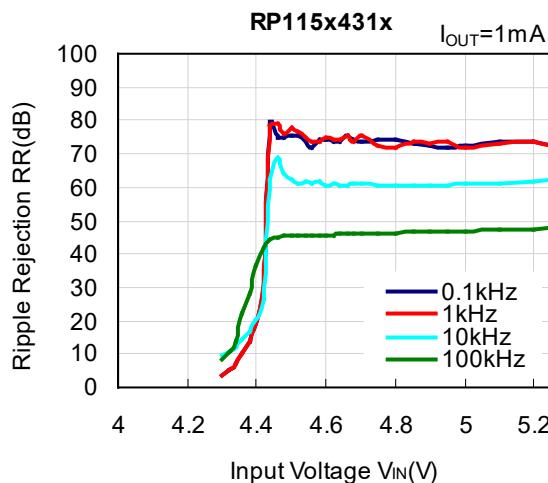
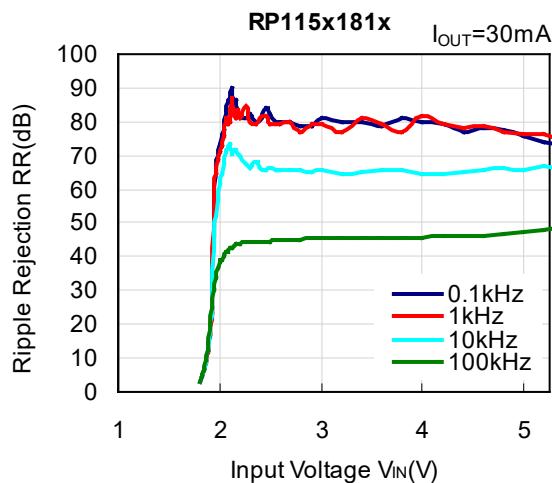
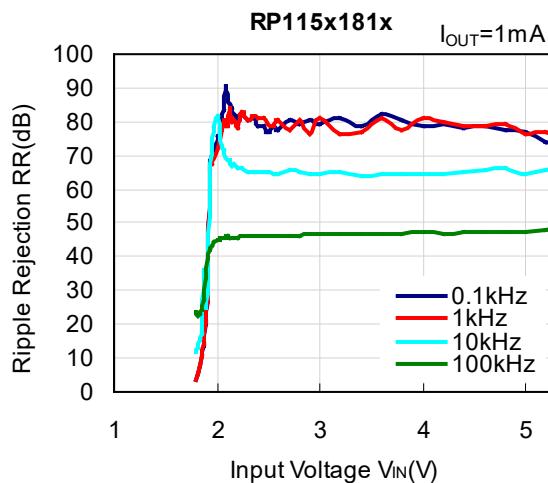
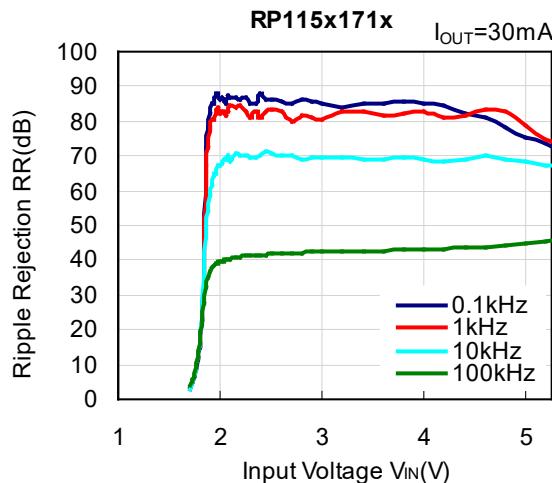
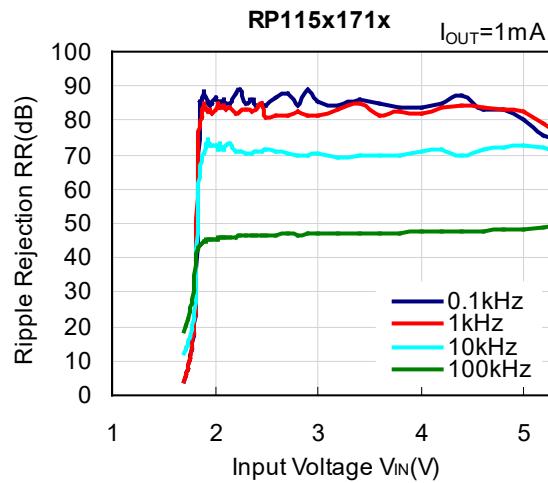


8) Dropout Voltage vs. Set Output Voltage ( $C_{IN}$  = Ceramic 1.0  $\mu$ F,  $C_{OUT}$  = Ceramic 1.0  $\mu$ F,  $T_a$  = 25°C)9) Dropout Voltage vs. Temperature ( $C_{IN}$  = Ceramic 1.0  $\mu$ F,  $C_{OUT}$  = Ceramic 1.0  $\mu$ F)

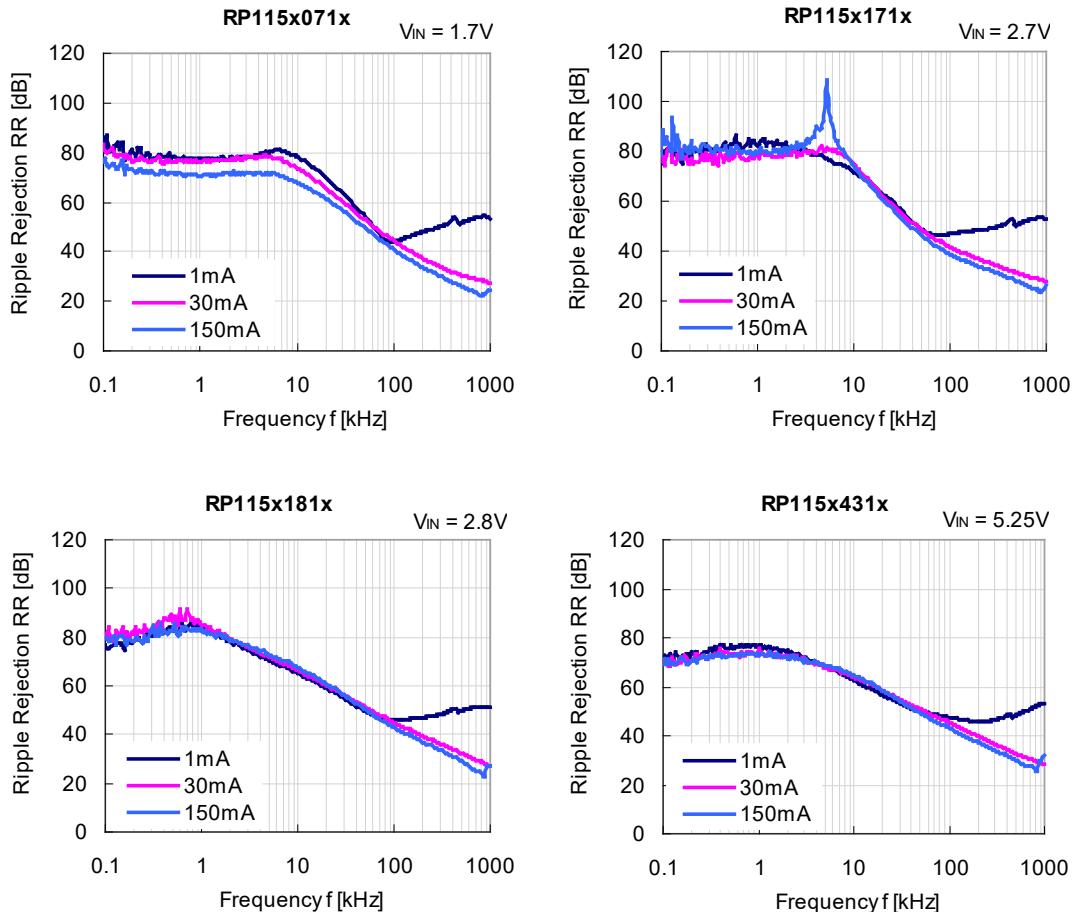


#### 10) Ripple Rejection vs. Input Voltage ( $C_{IN}$ = none, $C_{OUT}$ = Ceramic 1.0 $\mu$ F, Ripple = 0.2 Vp-p, $T_a$ = 25°C)

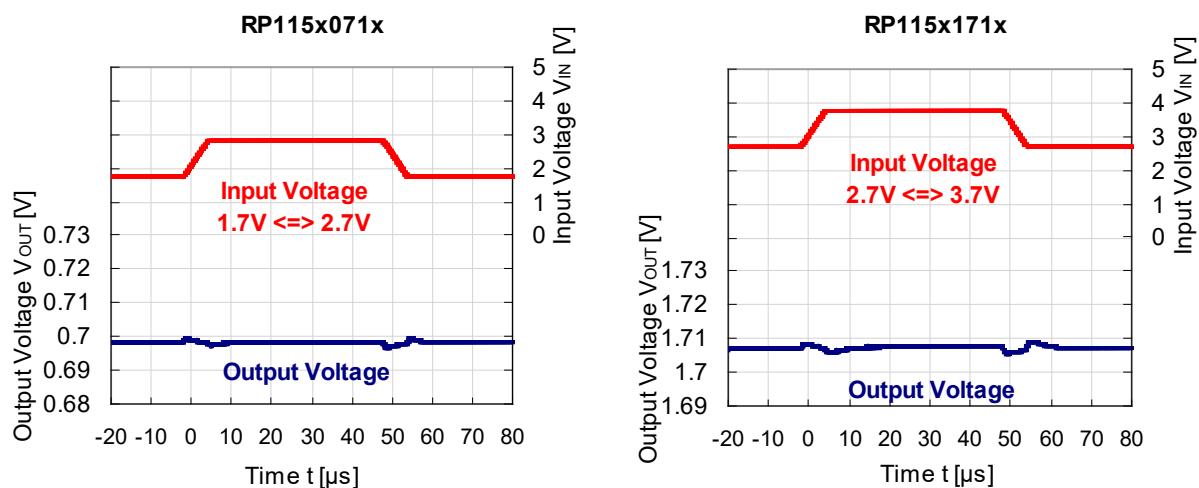


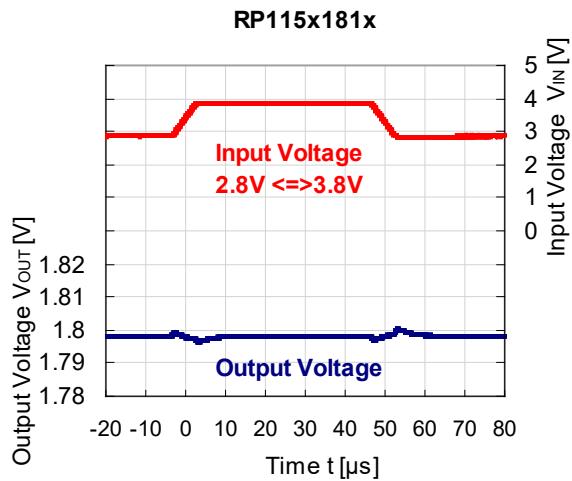


**11) Ripple Rejection vs. Frequency ( $C_{IN}$  = none,  $C_{OUT}$  = Ceramic 1.0  $\mu$ F, Ripple = 0.2 Vp-p, Ta = 25°C)**

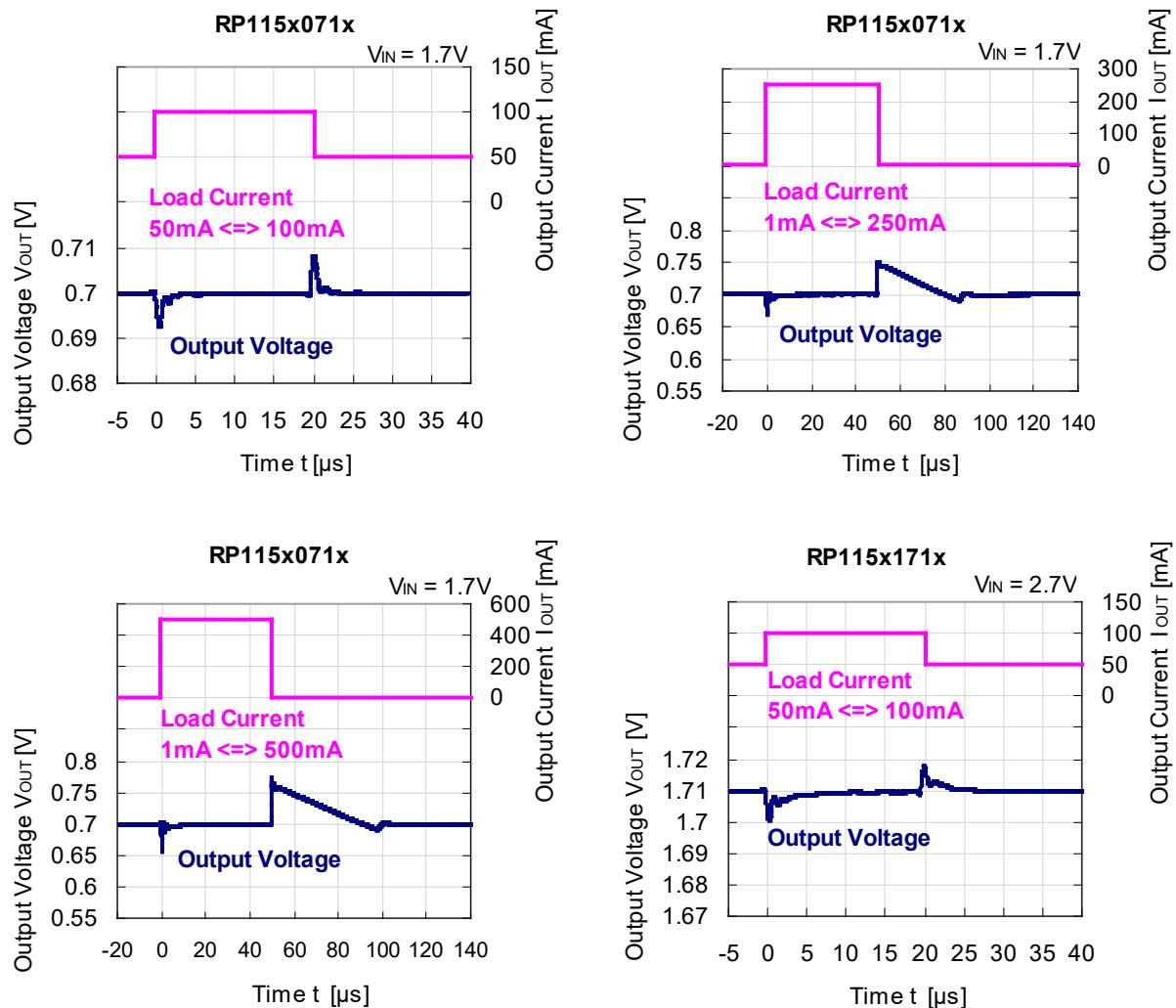


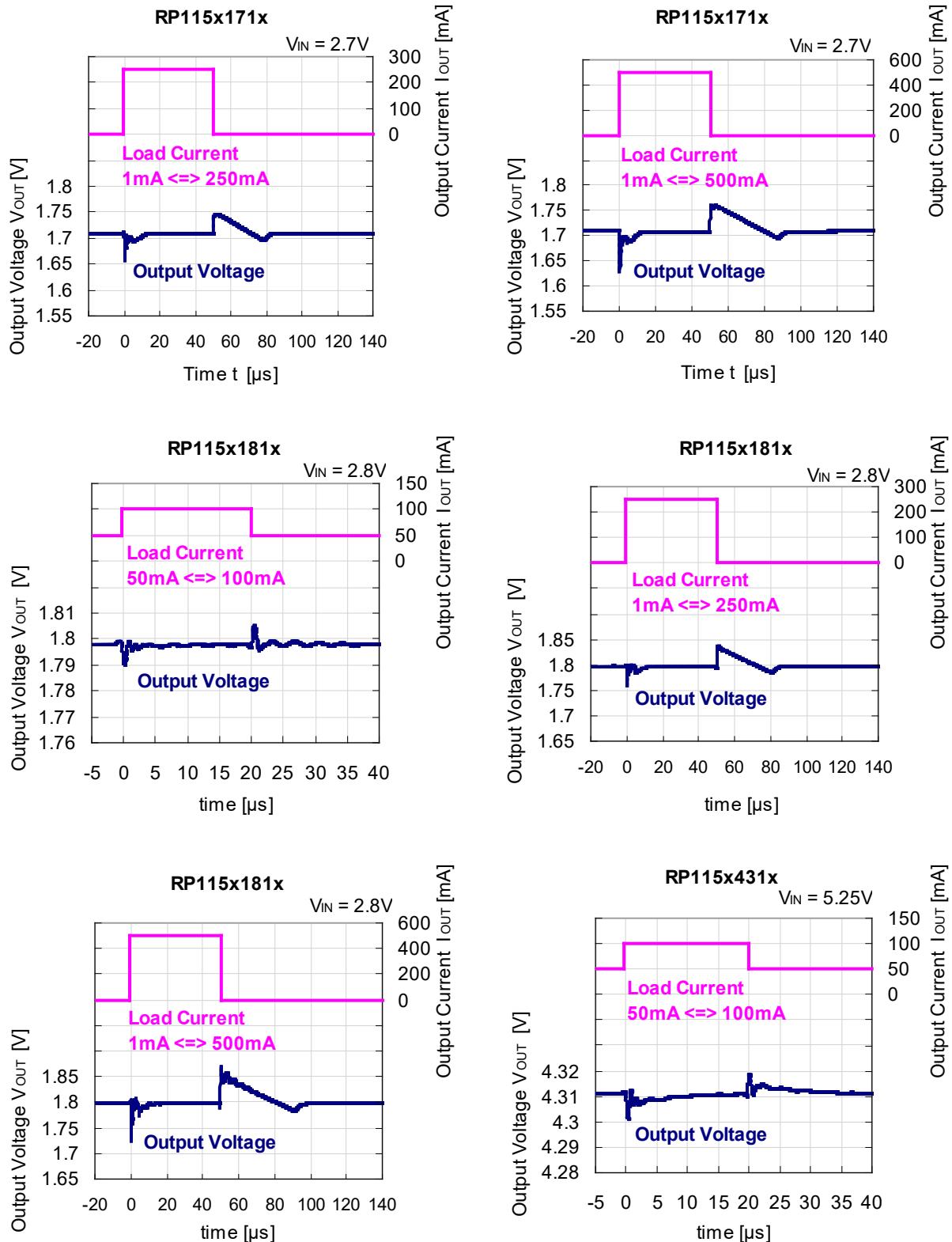
**12) Line Transient Response  $C_{IN}$  = none,  $C_{OUT}$  = Ceramic 1.0  $\mu$ F,  $I_{OUT}$  = 30 mA,  $t_r = t_f = 5 \mu$ s, Ta = 25°C)**

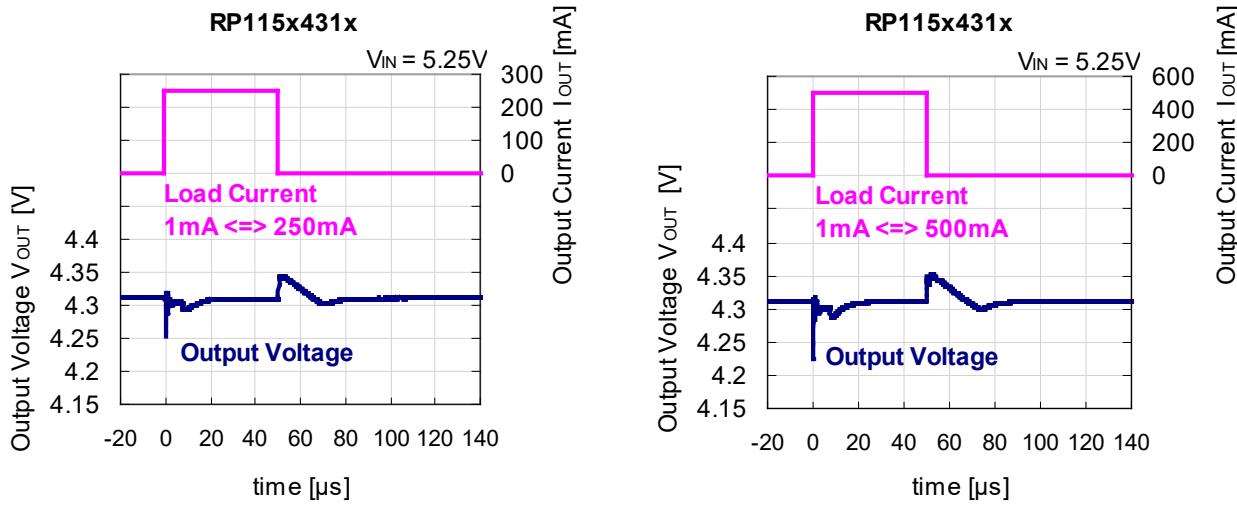




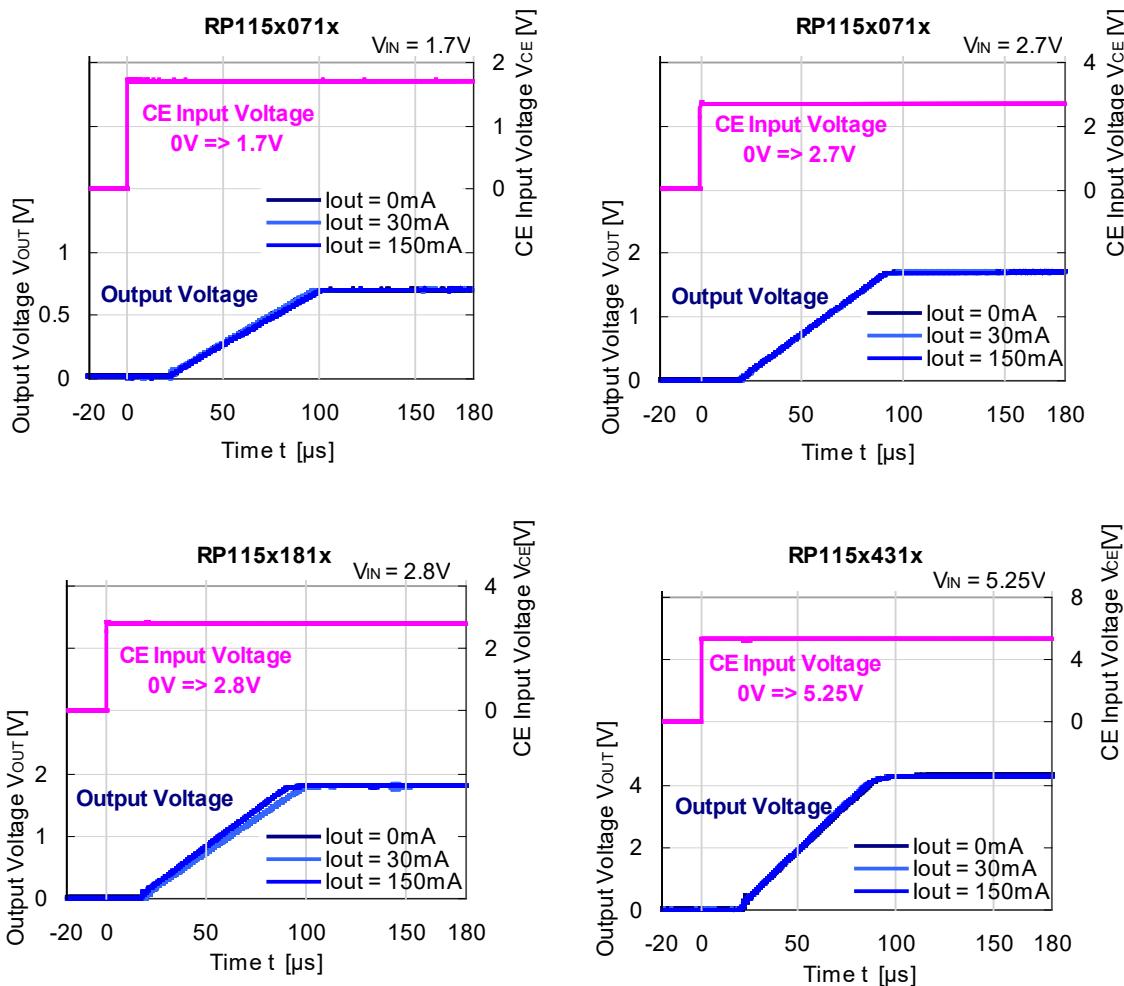
**13) Load Transient Response ( $C_{IN}$  = Ceramic 1.0  $\mu$ F,  $C_{OUT}$  = Ceramic 1.0  $\mu$ F,  $tr = tf = 0.5 \mu$ s,  $T_a = 25^\circ\text{C}$ )**

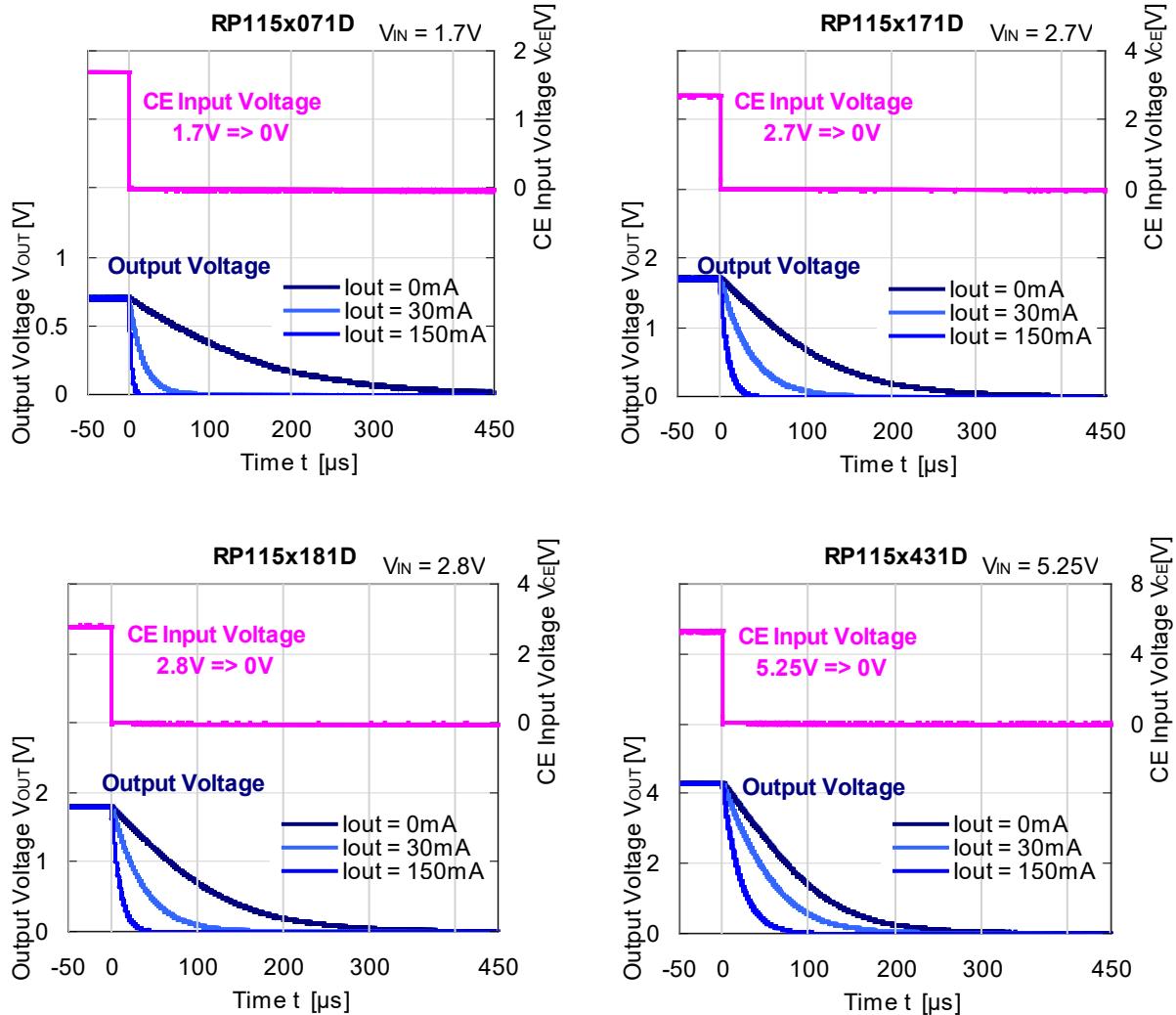




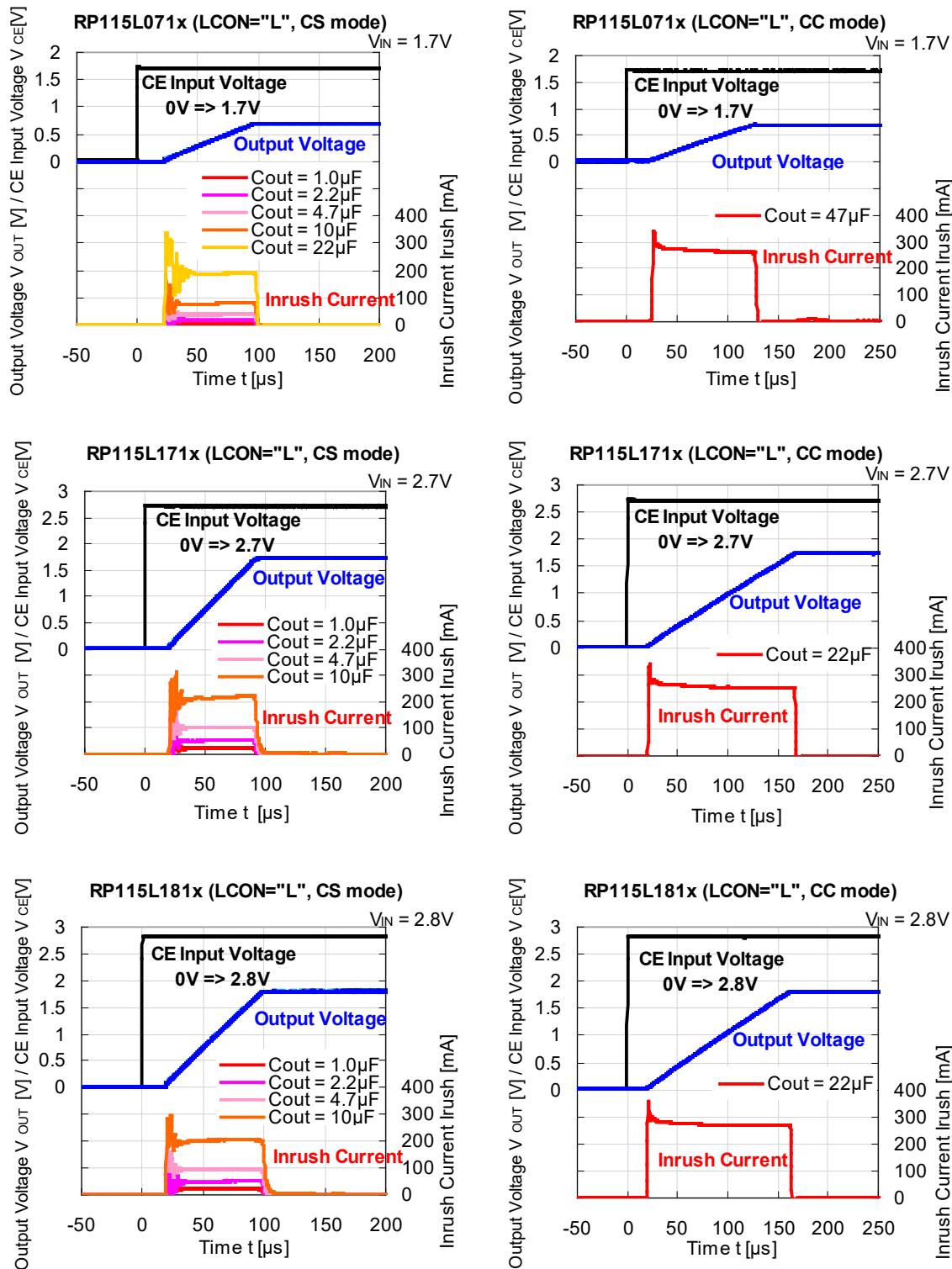


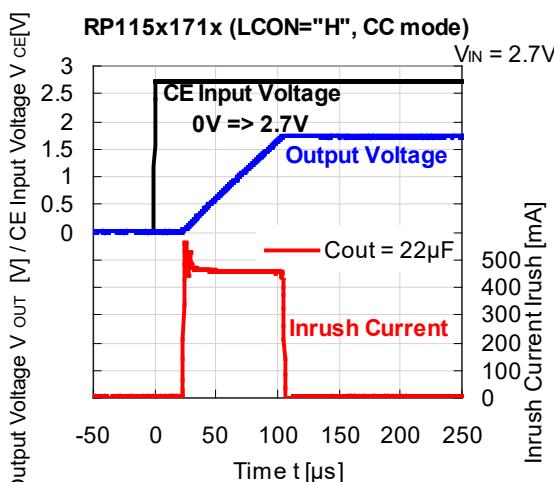
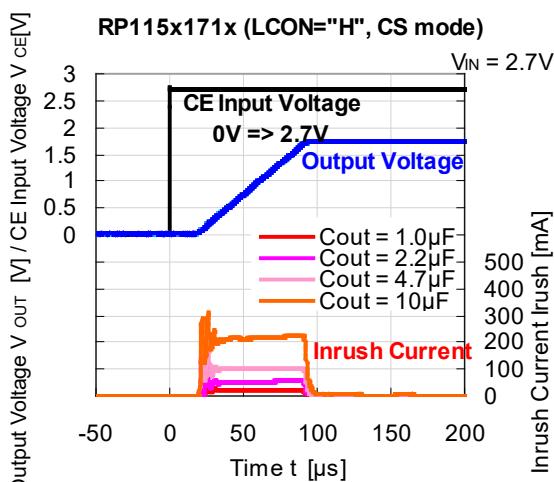
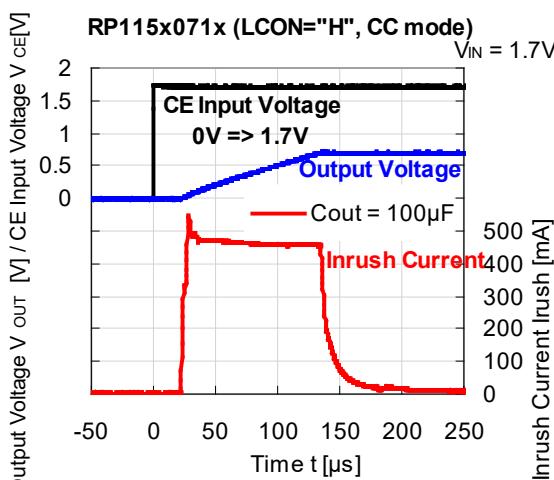
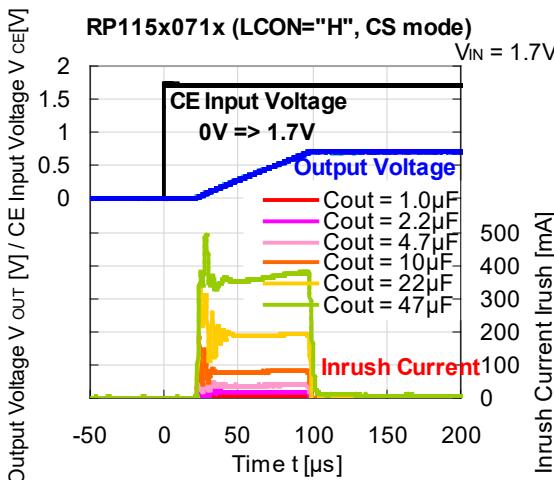
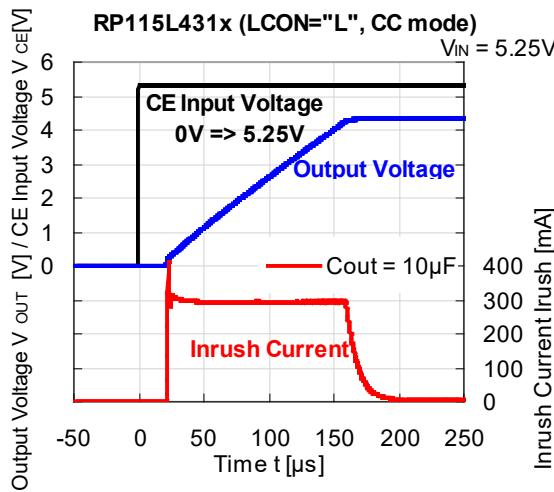
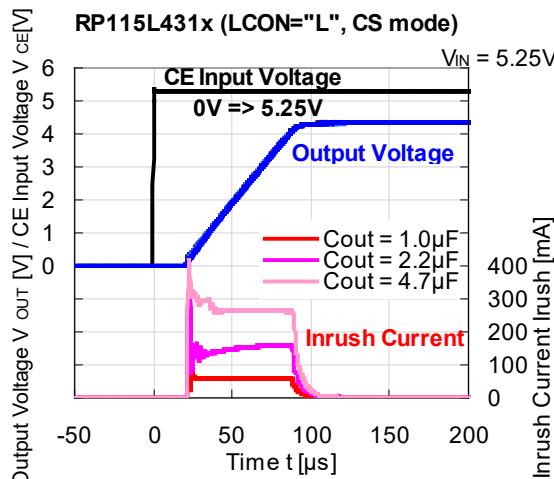
#### 14) Turn-on Waveform by CE Pin Signal ( $C_{IN} = \text{Ceramic } 1.0 \mu\text{F}$ , $C_{OUT} = \text{Ceramic } 1.0 \mu\text{F}$ , $T_a = 25^\circ\text{C}$ )

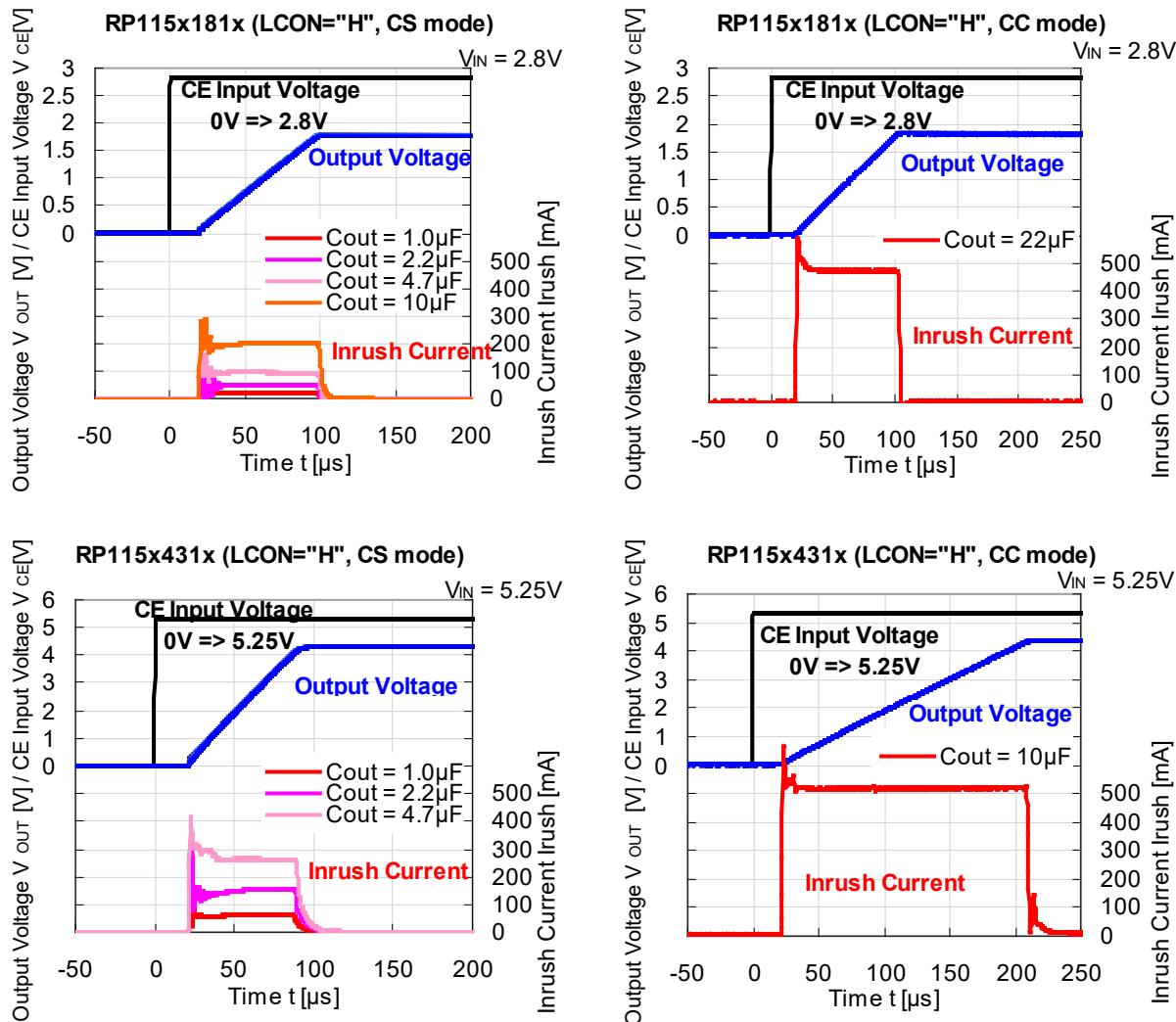


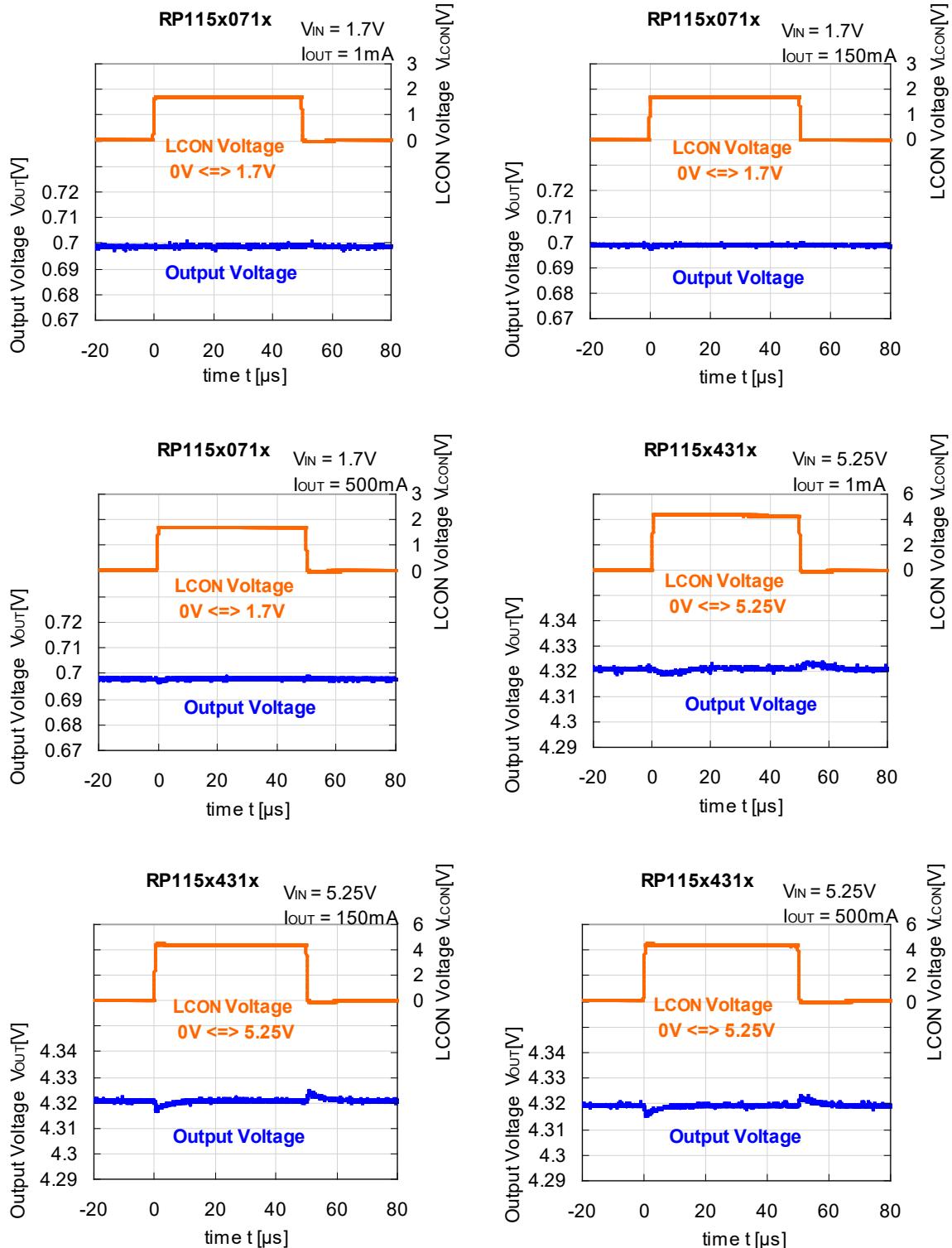
15) Turn-off Waveform by CE Pin Signal ( $C_{IN}$  = Ceramic 1.0  $\mu$ F,  $C_{OUT}$  = Ceramic 1.0  $\mu$ F, Ta = 25°C)

**16) Inrush Current ( $C_{IN}$  = Ceramic 1.0  $\mu$ F,  $I_{OUT} = 0$  mA,  $T_a = 25^\circ\text{C}$ )**







17) LCON Pin Transient Response ( $C_{IN}$  = Ceramic 1.0  $\mu$ F,  $C_{OUT}$  = Ceramic 1.0  $\mu$ F,  $T_a$  = 25°C)

The power dissipation of the package is dependent on PCB material, layout, and environmental conditions. The following conditions are based on JEDEC STD. 51-7.

### Measurement Conditions

Item	Measurement Conditions
Environment	Mounting on Board (Wind Velocity = 0 m/s)
Board Material	Glass Cloth Epoxy Plastic (Four-Layer Board)
Board Dimensions	76.2 mm × 114.3 mm × 0.8 mm
Copper Ratio	Outer Layer (First Layer): Less than 95% of 50 mm Square Inner Layers (Second and Third Layers): Approx. 100% of 50 mm Square Outer Layer (Fourth Layer): Approx. 100% of 50 mm Square
Through-holes	Ø 0.2 mm × 15 pcs

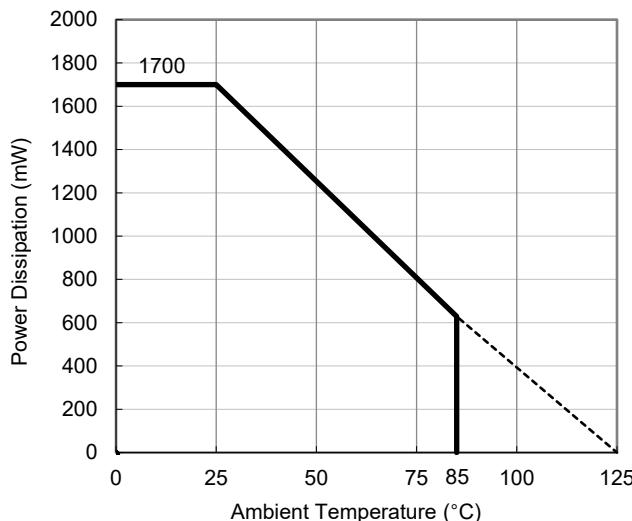
### Measurement Result

(Ta = 25°C, Tjmax = 125°C)

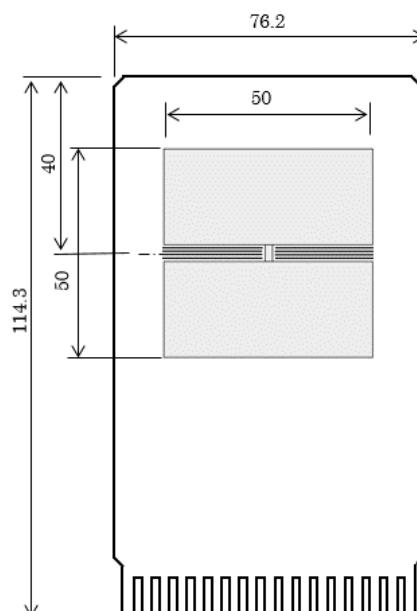
Item	Measurement Result
Power Dissipation	1700 mW
Thermal Resistance ( $\theta_{ja}$ )	$\theta_{ja} = 56^\circ\text{C}/\text{W}$
Thermal Characterization Parameter ( $\psi_{jt}$ )	$\psi_{jt} = 18^\circ\text{C}/\text{W}$

$\theta_{ja}$ : Junction-to-Ambient Thermal Resistance

$\psi_{jt}$ : Junction-to-Top Thermal Characterization Parameter



Power Dissipation vs. Ambient Temperature

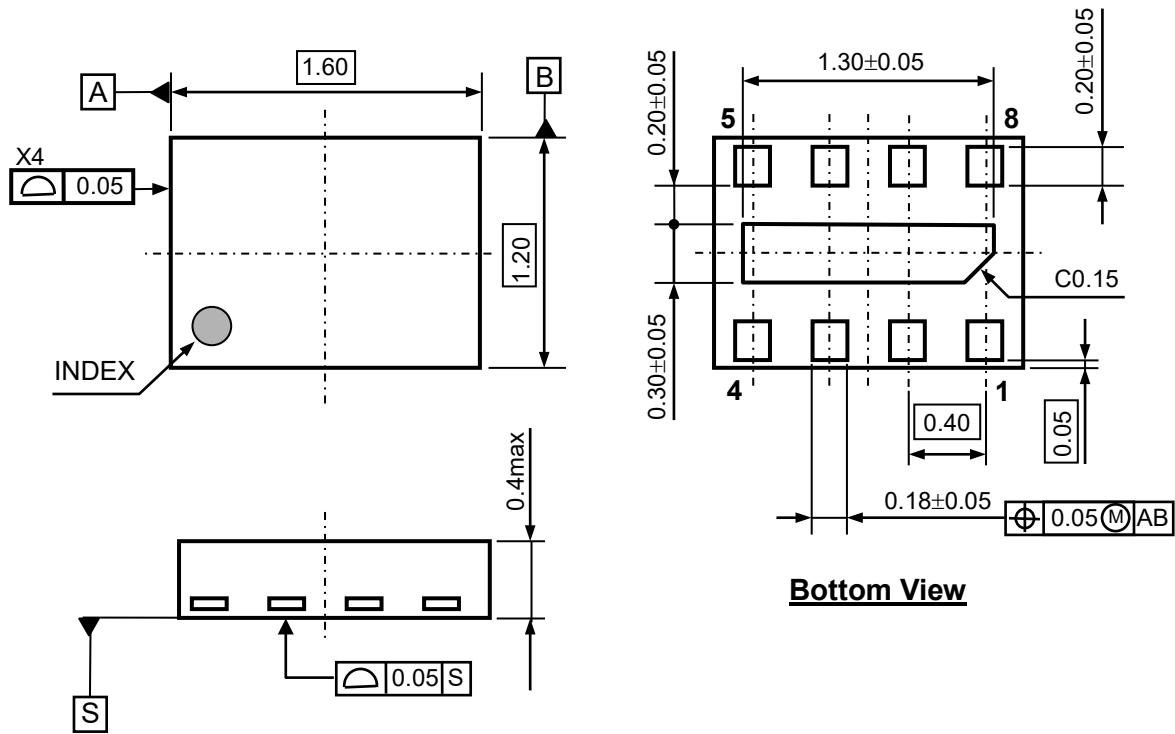


Measurement Board Pattern

## PACKAGE DIMENSIONS

DFN1216-8

DM-DFN1216-8-JE-B



UNIT: mm

DFN1216-8 Package Dimensions

# POWER DISSIPATION

SOT-89-5

Ver. A

The power dissipation of the package is dependent on PCB material, layout, and environmental conditions. The following measurement conditions are based on JEDEC STD. 51-7.

## Measurement Conditions

Item	Measurement Conditions
Environment	Mounting on Board (Wind Velocity = 0 m/s)
Board Material	Glass Cloth Epoxy Plastic (Four-Layer Board)
Board Dimensions	76.2 mm × 114.3 mm × 0.8 mm
Copper Ratio	Outer Layer (First Layer): Less than 95% of 50 mm Square Inner Layers (Second and Third Layers): Approx. 100% of 50 mm Square Outer Layer (Fourth Layer): Approx. 100% of 50 mm Square
Through-holes	φ 0.3 mm × 13 pcs

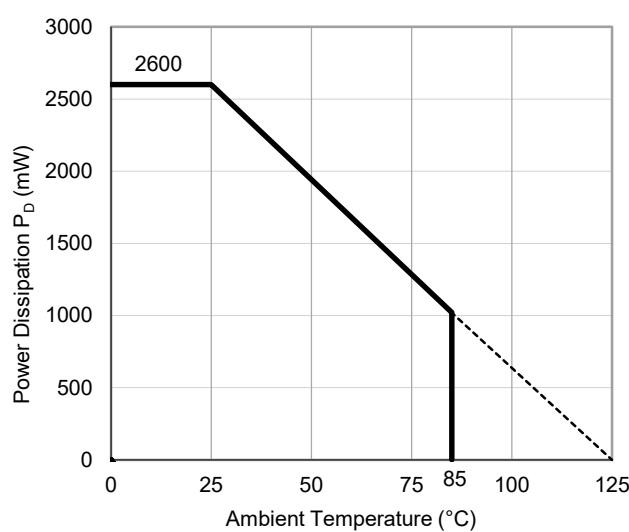
## Measurement Result

(Ta = 25°C, Tjmax = 125°C)

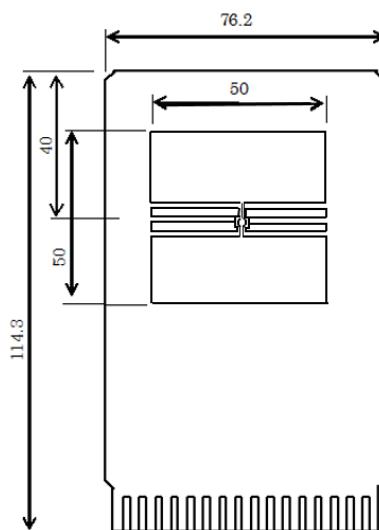
Item	Measurement Result
Power Dissipation	2600 mW
Thermal Resistance ( $\theta_{ja}$ )	$\theta_{ja} = 38^\circ\text{C}/\text{W}$
Thermal Characterization Parameter ( $\psi_{jt}$ )	$\psi_{jt} = 13^\circ\text{C}/\text{W}$

$\theta_{ja}$ : Junction-to-Ambient Thermal Resistance

$\psi_{jt}$ : Junction-to-Top Thermal Characterization Parameter



Power Dissipation vs. Ambient Temperature

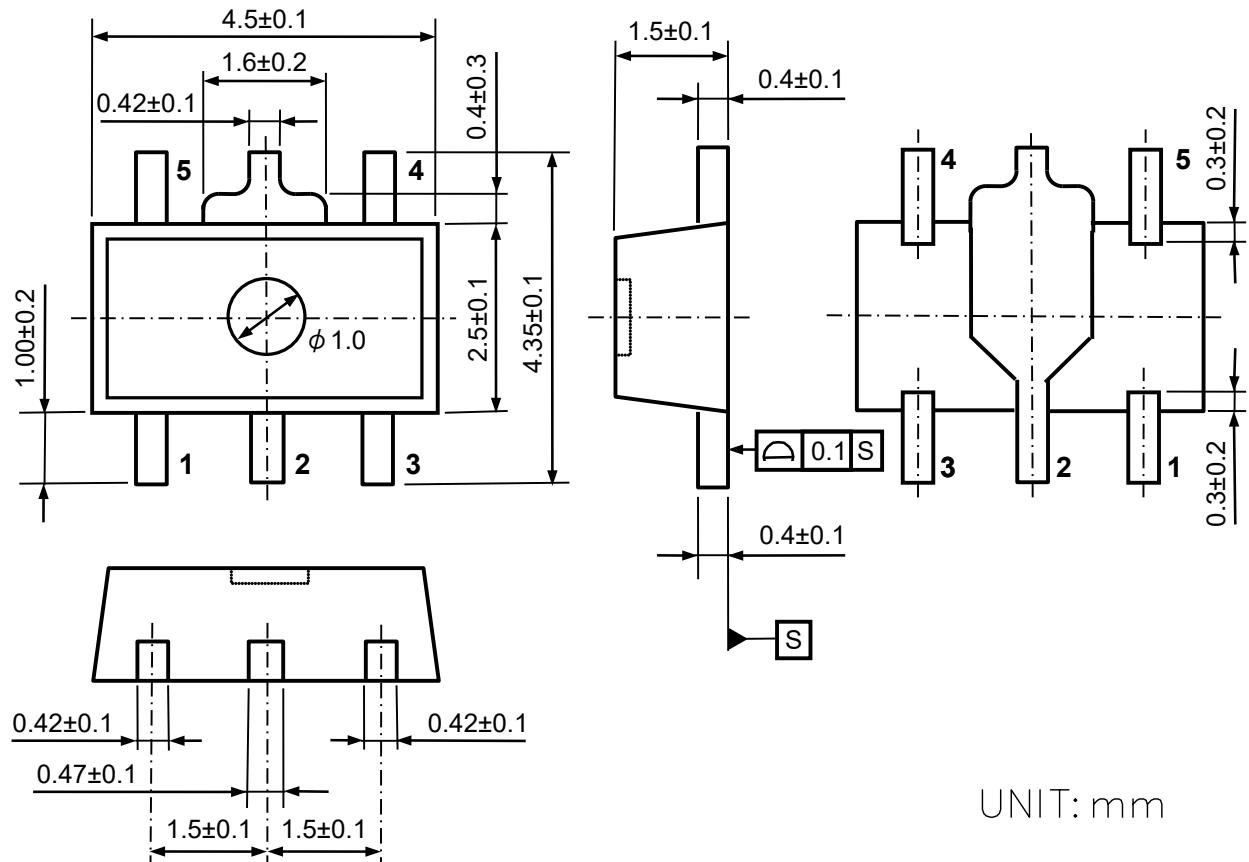


Measurement Board Pattern

## PACKAGE DIMENSIONS

SOT-89-5

DM-SOT-89-5-JE-B



SOT-89-5 Package Dimensions



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